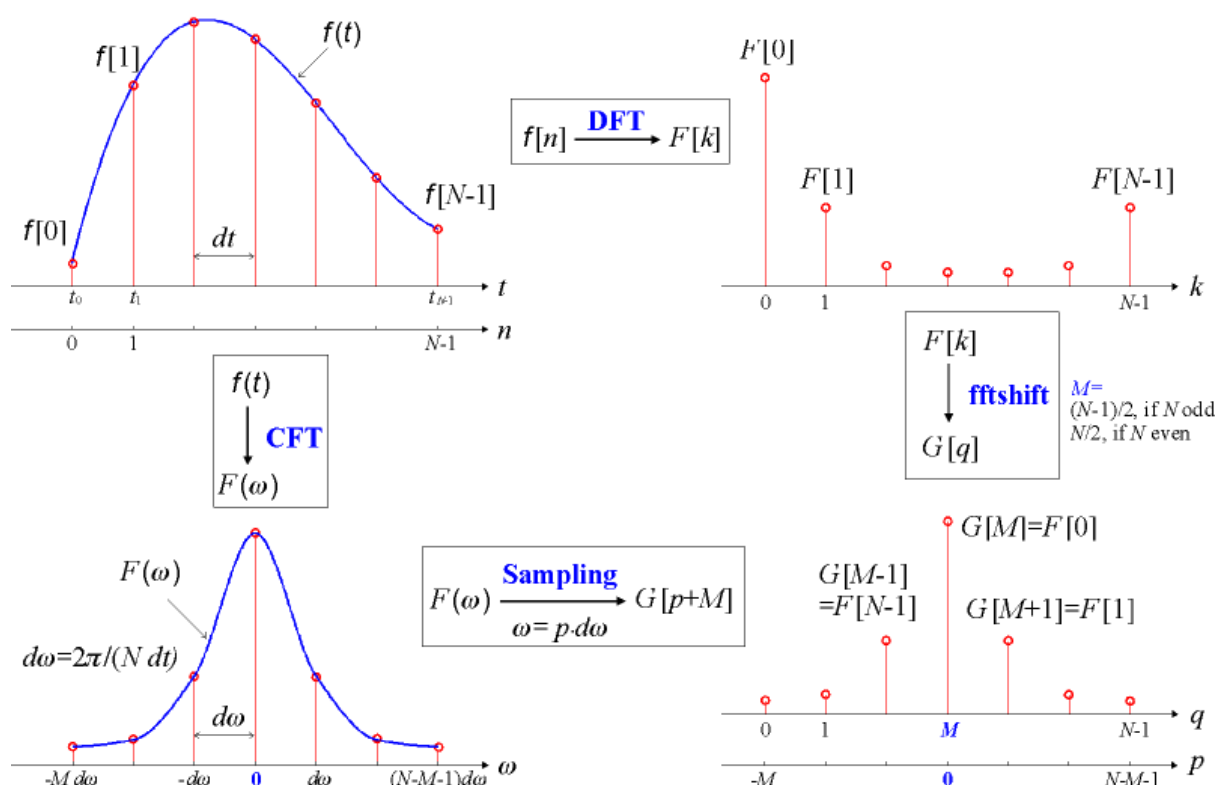


How to simulate CFT by DFT(or FFT)?

■ Definitions

- 1) CFT: $F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$
- 2) DFT: $F[k] = \sum_{n=0}^{N-1} f[n] \cdot e^{-j\frac{2\pi}{N}k \cdot n}$, $k = \{0, 1, \dots, N-1\}$.

■ Schematic



■ Special cases

- 1) $F(\omega=0) = \int_{-\infty}^{\infty} f(t)dt$, by trapezoidal approximation, $\approx \left(\sum_{n=0}^{N-1} f[n] \right) \cdot dt$;

$$F[k=0] = \sum_{n=0}^{N-1} f[n]; \Rightarrow \boxed{F(\omega=0) = dt \cdot F[k=0]}.$$

- (a) The absolute sampling times play no role.
- (b) DC spectrum is always sampled.

- 2) Let spectral resolution $d\omega = \frac{2\pi}{N \cdot dt}$. $F(\omega=d\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$, by $t = t_0 + n \cdot dt$ and

trapezoidal approximation, $\approx \sum_{n=0}^{N-1} f[n] \cdot e^{-j\omega \cdot (t_0 + n \cdot dt)} \cdot dt = dt \cdot e^{-j\omega \cdot t_0} \cdot \left(\sum_{n=0}^{N-1} f[n] \cdot e^{-j\frac{2\pi}{N}n} \right);$

$$F[k=1] = \sum_{n=0}^{N-1} f[n] \cdot e^{-j\frac{2\pi}{N}n}; \Rightarrow \boxed{F(\omega=\omega_0) = dt \cdot e^{-j\omega_0 \cdot t_0} \cdot F[k=1]}.$$

Now the absolute sampling times matter.

3) $F(\omega=-\omega_0) = \int_{-\infty}^{\infty} f(t) e^{j\omega_0 t} dt$, by $t=t_0+n \cdot dt$ and trapezoidal approximation,

$$\approx \sum_{n=0}^{N-1} f[n] \cdot e^{j\omega_0 (t_0 + n \cdot dt)} \cdot dt = dt \cdot e^{j\omega_0 \cdot t_0} \cdot \left(\sum_{n=0}^{N-1} f[n] \cdot e^{j\frac{2\pi}{N}n} \right);$$

$$F[k=N-1] = \sum_{n=0}^{N-1} f[n] \cdot e^{-j\frac{2\pi}{N}(N-1)n} = \sum_{n=0}^{N-1} f[n] \cdot e^{j\frac{2\pi}{N}n}; \Rightarrow \boxed{F(\omega=-\omega_0) = dt \cdot e^{j\omega_0 \cdot t_0} \cdot F[k=N-1]}.$$

■ Conclusion

Since $\text{fftshift}\{F[k]\} = G[q]$, $q = \{0, 1, \dots, M, \dots, N-1\}$;

where $G[q=M] = F[k=0]$, and $\boxed{M = \begin{cases} (N-1)/2, & \text{if } N \in \text{odd} \\ N/2, & \text{if } N \in \text{even} \end{cases}};$

$$\Rightarrow \boxed{F(\omega=\omega_0 + \Delta\omega) = dt \cdot e^{-j\omega_0 \cdot t_0} \cdot G[q=p+M]};$$

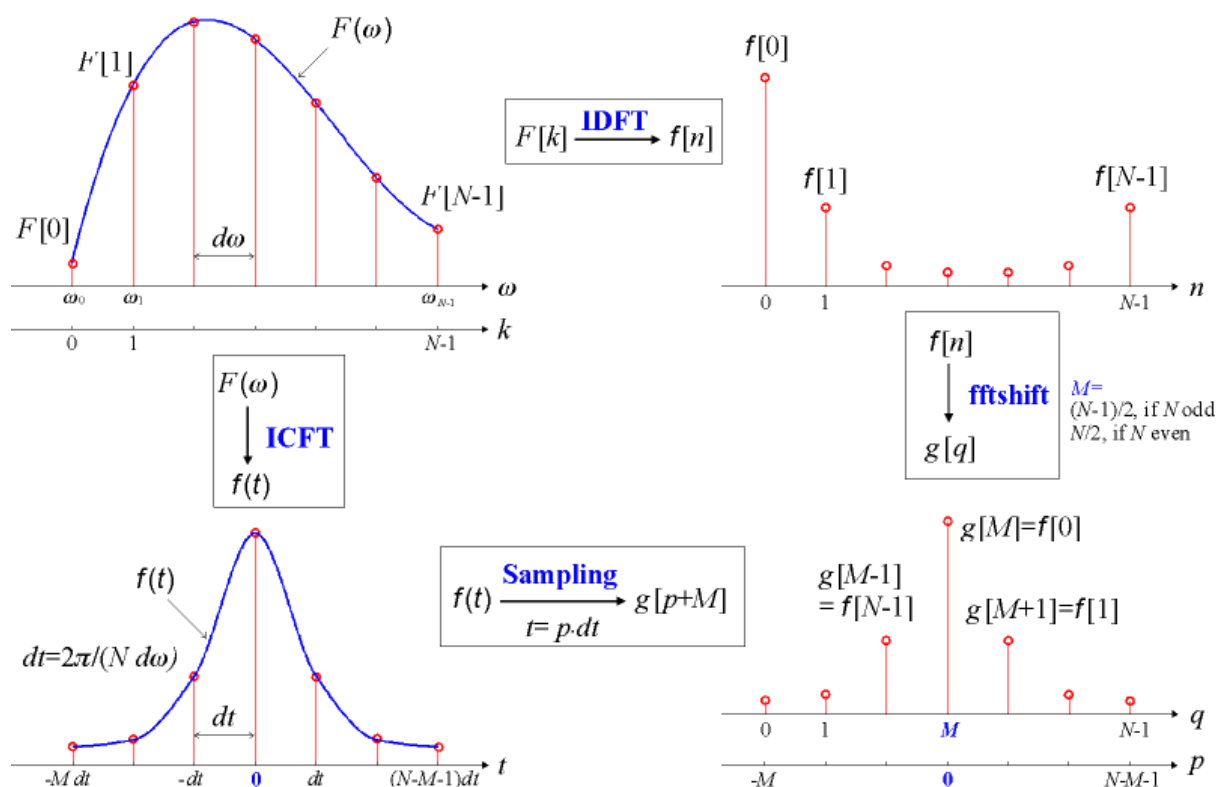
- 1) A multiplicative factor **dt** is required.
- 2) DFT is unaware of absolute time (but index 0, 1, ...). It always **assumes an initial time of $t=0$** , which differs from the actual value **$t=t_0$** . Consequently, a **linear spectral phase** term $e^{-j\omega \cdot t_0}$ is used to **compensate** this fake delay.

How to simulate ICFT by IDFT(or IFFT)?

■ Definitions

- 1) ICFT: $f(t) = \frac{1}{2p} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$
- 2) DFT: $f[n] = \frac{1}{N} \sum_{k=0}^{N-1} F[k] \cdot e^{-j\frac{2p}{N} n \cdot k}$, $n = \{0, 1, \dots, N-1\}$.

■ Schematic



■ Special cases

- 1) $f(t=0) = \frac{1}{2p} \int_{-\infty}^{\infty} F(\omega) d\omega$, by trapezoidal approximation, $\approx \frac{1}{2p} \left(\sum_{k=0}^{N-1} F[k] \right) \cdot d\omega$;

$$f[n=0] = \frac{1}{N} \sum_{k=0}^{N-1} F[k]; \Rightarrow \boxed{f(t=0) = \frac{N \cdot d\omega}{2p} \cdot f[n=0]}.$$

- (a) The absolute sampling frequencies play no role.
- (b) Zero-time is always sampled.

- 2) Let temporal resolution $dt = \frac{2p}{N \cdot d\omega}$. $f(t=dt) = \frac{1}{2p} \int_{-\infty}^{\infty} F(\omega) e^{j\omega \cdot dt} d\omega$, by $\omega = \omega_0 + k \cdot d\omega$ and

$$\text{trapezoidal approximation, } \approx \frac{1}{2p} \sum_{k=0}^{N-1} F[k] \cdot e^{j(w_0+k \cdot dw)dt} \cdot dw = \frac{dw}{2p} \cdot e^{jw_0 \cdot dt} \cdot \left(\sum_{k=0}^{N-1} F[k] \cdot e^{j\frac{2p}{N}k} \right);$$

$$f[n=1] = \frac{1}{N} \sum_{k=0}^{N-1} F[k] \cdot e^{-j\frac{2p}{N}k}; \Rightarrow \boxed{f(t=dt) = \frac{N \cdot dw}{2p} \cdot e^{jw_0 \cdot dt} \cdot f[n=1]}.$$

Now the absolute sampling frequencies matter.

$$3) f(t=-dt) = \frac{1}{2p} \int_{-\infty}^{\infty} F(w) e^{-jw \cdot dt} dw, \text{ by } w=w_0+k \cdot dw \text{ and trapezoidal approximation,}$$

$$\approx \frac{1}{2p} \sum_{k=0}^{N-1} F[k] \cdot e^{-j(w_0+k \cdot dw)dt} \cdot dw = \frac{dw}{2p} \cdot e^{-jw_0 \cdot dt} \cdot \left(\sum_{k=0}^{N-1} F[k] \cdot e^{-j\frac{2p}{N}k} \right);$$

$$f[n=N-1] = \frac{1}{N} \sum_{k=0}^{N-1} F[k] \cdot e^{-j\frac{2p}{N}(N-1)k} = \frac{1}{N} \sum_{k=0}^{N-1} F[k] \cdot e^{j\frac{2p}{N}k};$$

$$\Rightarrow \boxed{f(t=-dt) = \frac{N \cdot dw}{2p} \cdot e^{-jw_0 \cdot dt} \cdot f[n=N-1]}.$$

■ Conclusion

Since $\text{fftshift}\{f[n]\} = g[q]$, $q = \{0, 1, \dots, M, \dots, N-1\}$;

where $g[q=M] = f[n=0]$, and $M = \begin{cases} (N-1)/2, & \text{if } N \in \text{odd} \\ N/2, & \text{if } N \in \text{even} \end{cases}$;

$$\Rightarrow \boxed{f(t=p \cdot dt)} = \boxed{\frac{N \cdot dw}{2p} \cdot e^{jw_0 \cdot p \cdot dt} \cdot g[q=p+M]};$$

1) A multiplicative factor $\boxed{\frac{N \cdot dw}{2p} = \frac{1}{dt}}$ is required.

2) IDFT is unaware of absolute frequency (but index 0, 1, ...). It always **assumes an initial angular frequency of $w=0$** , which differs from the actual value $w=w_0$. Consequently, a **linear temporal phase** term $e^{jw_0 \cdot t}$ is used to **compensate** this fake frequency shift.