

Chapter 7

Laplace Transform (LT)

- ❑ Definition, LT of basic functions (7.1)
- ❑ Inverse LT (7.2.1)
- ❑ Operational properties (7.2-7.4)
- ❑ Dirac delta function (7.5)

□ Definition, LT of basic functions

- Definition
- LT of basic functions
- Existence of LT

What's “function”?

3

- A “relation” mapping input **value(s)** to output **value(s)**, just like a “machine” mapping raw materials to products.
- E.g. $f(x) = \sin(x)$, mapping an input value x_0 to an output value $\sin(x_0)$.
- E.g. $f(x,y) = x + y$, mapping a pair of input values (x_0, y_0) to an output value $x_0 + y_0$.
- E.g. $f(z) = z^*$, mapping a complex number z_0 (a pair of real numbers) to a complex number $(z_0)^*$ (a pair of real numbers).

What's “transform”?

4

- A “relation” mapping an input **function** to an output, **function**.
- E.g. differentiation $D\{f(x)\} = f'(x)$, mapping an input function $f(x)$ to an output function $f'(x)$.
- E.g. partial differentiation $D_x\{f(x,y)\} = \partial f(x,y)/\partial x$, mapping a 2-variable function $f(x,y)$ to another 2-variable function $\partial f(x,y)/\partial x$.
- It's **linear** if $T\{a \times f_1(x) + b \times f_2(x)\} = a \times T\{f_1(x)\} + b \times T\{f_2(x)\}$.

What's “integral transform”?

5

- The output function $F(s)$ is obtained by integrating the product of the input function $f(t)$ and a **kernal function** $K(s,t)$:

$$F(s) \equiv T\{f(t)\} = \int_0^{\infty} f(t) \times K(s,t) dt$$

- Note: Output function has a new independent variable s .
- All integral transforms are **linear**. (Why?)

What's Laplace transform (LT)?

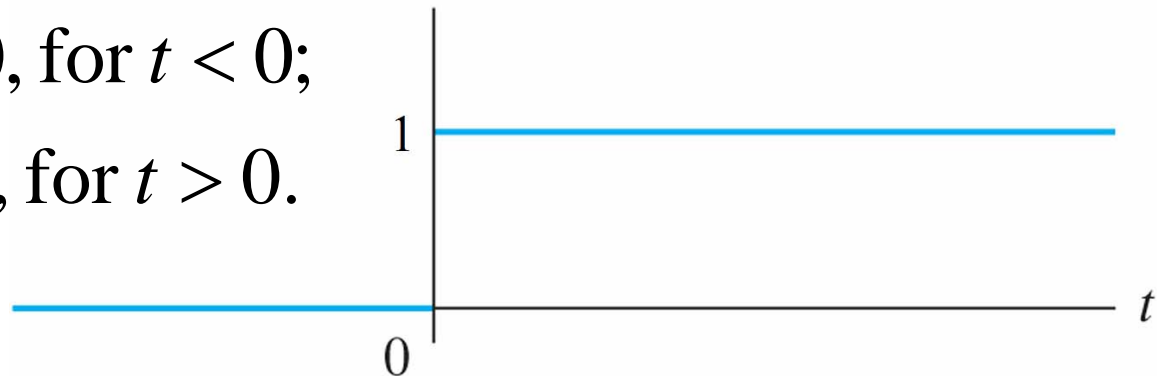
6

- A special case of integral transform where the kernel is an exponential function $K(s,t) = e^{-st}$:

$$F(s) \equiv L\{f(t)\} = \int_0^{\infty} f(t) \times e^{-st} dt.$$

- Integration is carried out over $0 < t < \infty$, $\Rightarrow L\{f(t)\} = L\{f(t) \times u(t)\}$, where the unit-step function is:

$$u(t) \equiv \begin{cases} 0, & \text{for } t < 0; \\ 1, & \text{for } t > 0. \end{cases}$$



Example 1: $L\{1\}$

7

- By definition,

$$\begin{aligned} L\{1\} &= L\{u(t)\} = \int_0^{\infty} u(t) \times e^{-st} dt = \int_0^{\infty} 1e^{-st} dt \\ &= \left. \frac{e^{-st}}{-s} \right|_0^{\infty} = \frac{0-1}{-s} = \frac{1}{s}, \quad \text{if } s > 0. \end{aligned}$$

- $L\{1\}$ exists (integrable) only in a finite region of convergence ROC: $\{s > 0\}$.

Example 2: $L\{t\}$

8

- By definition,

$$L\{t\} = L\{t \times u(t)\} = \int_0^{\infty} t \times u(t) \times e^{-st} dt = \int_0^{\infty} te^{-st} dt;$$

- Integration by parts: $u = t, v' = e^{-st}, \Rightarrow u'=1, v = -e^{-st}/s$

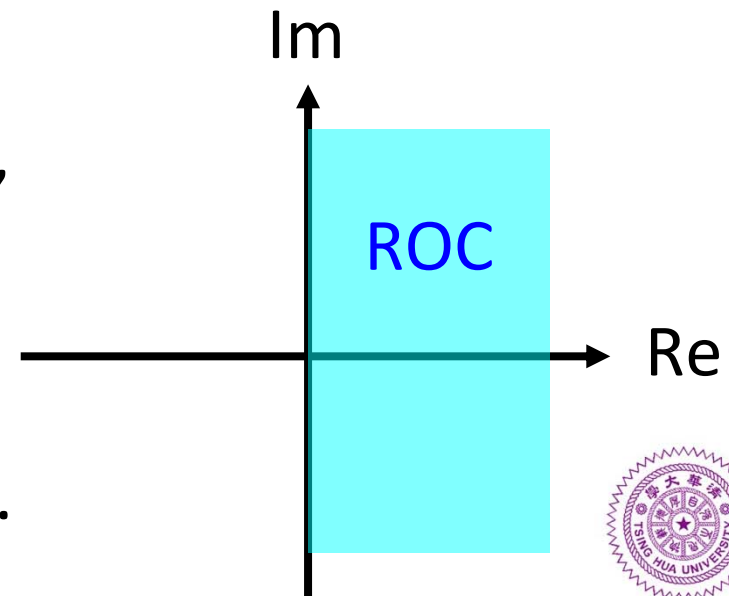
$$L\{t\} = uv \Big|_0^{\infty} - \int_0^{\infty} u'v dt = -\frac{te^{-st}}{s} \Big|_0^{\infty} + \int_0^{\infty} \frac{e^{-st}}{s} dt$$

$$= -\left(\frac{0-0}{s}\right) - \left(\frac{e^{-st}}{s^2} \Big|_0^{\infty}\right) = -\frac{0-1}{s^2} = \frac{1}{s^2}, \text{ if } s > 0$$

Complex nature

9

- The new variable s is actually complex ($s = \alpha + j\beta \in \mathbb{C}$),
 $\Rightarrow K(s, t) = e^{-st} = e^{-(\alpha + j\beta)t} = e^{-\alpha t} \times [\cos(\beta t) + j\sin(\beta t)] \in \mathbb{C}$,
 $\Rightarrow F(s) = \int_0^\infty f(t) \times e^{-st} dt \in \mathbb{C}$
- Input $f(t)$: a real function of real variable t .
- Output $F(s)$: a complex function of complex variable s ,
 $\Rightarrow \text{Re}\{F\}, \text{Im}\{F\}$ are functions of 2 real variables.



ROC: $\{s > 0\}$ is actually $\text{Re}\{s\} > 0$.

Example: $L\{t^2\}$

10

- By definition, $L\{t^2\} = \int_0^\infty t^2 e^{-st} dt$;
- Let $\{u = t^2, v' = e^{-st}\}$, $\Rightarrow \{u' = 2t, v = -e^{-st}/s\}$

$$\begin{aligned} L\{t^2\} &= uv \Big|_0^\infty - \int_0^\infty u'v dt = -\frac{t^2 e^{-st}}{s} \Big|_0^\infty + \frac{2}{s} \int_0^\infty t e^{-st} dt \\ &= 0 + \frac{2}{s} L\{t\} = \frac{2}{s} \times \frac{1}{s^2} = \frac{2}{s^3} = \frac{2!}{s^{2+1}}, \text{ if } \operatorname{Re}\{s\} > 0 \end{aligned}$$

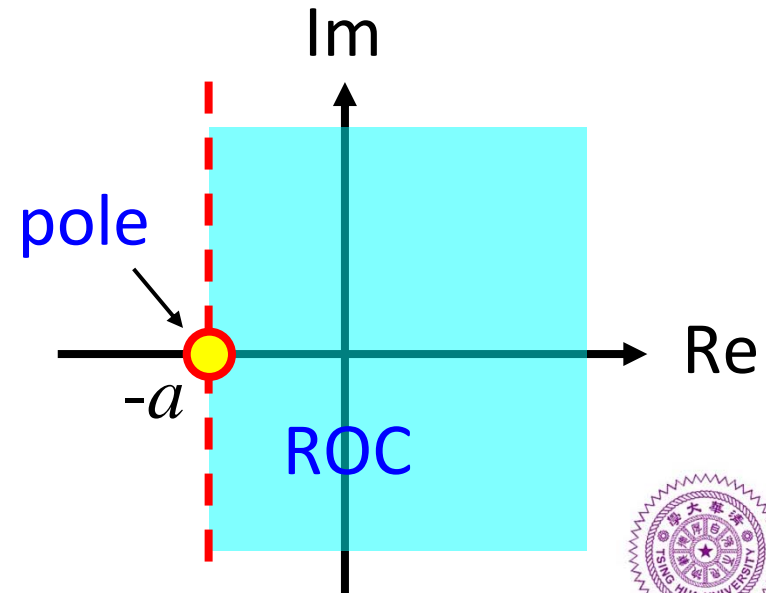
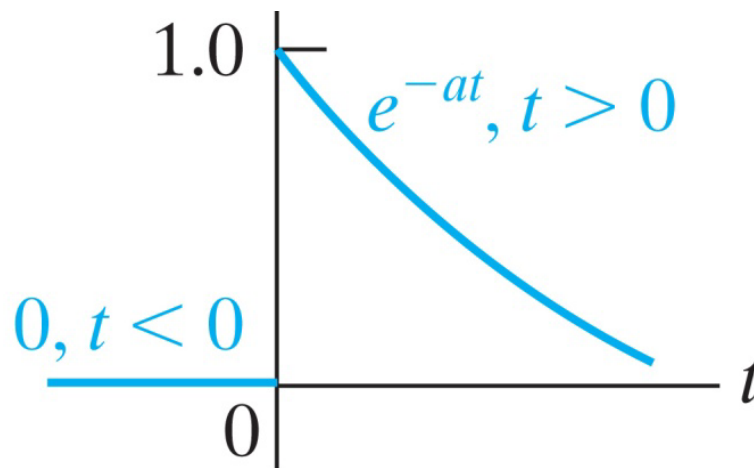
- Recursive relation gives $L\{t^n\} = n!/s^{n+1}$.

Example 3: $L\{e^{-at}\}$

11

- By definition,

$$\begin{aligned} L\{e^{-at}\} &= \int_0^{\infty} e^{-at} e^{-st} dt = \int_0^{\infty} e^{-(a+s)t} dt \\ &= \frac{e^{-(a+s)t}}{-(a+s)} \Big|_0^{\infty} = \frac{1}{s+a}, \text{ if } \operatorname{Re}\{s\} > -a. \end{aligned}$$



Example 4: $L\{\sin(\omega t)\}$

12

- By definition,

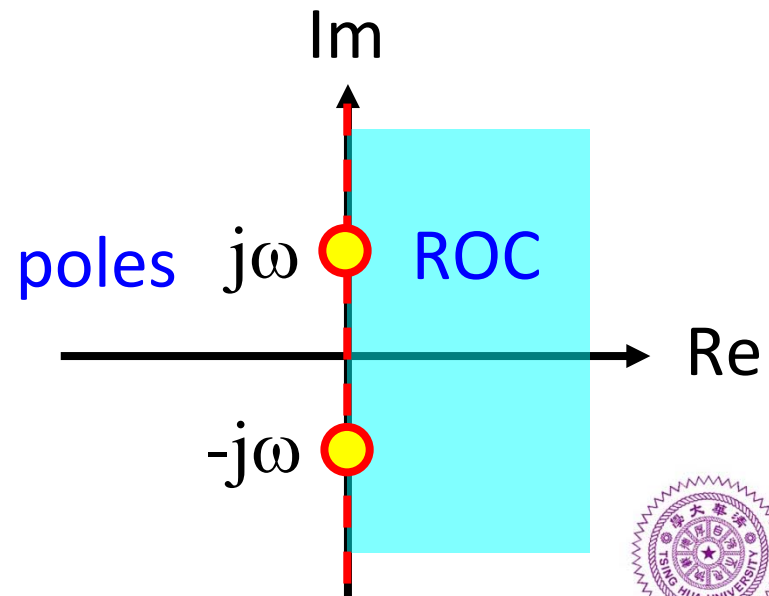
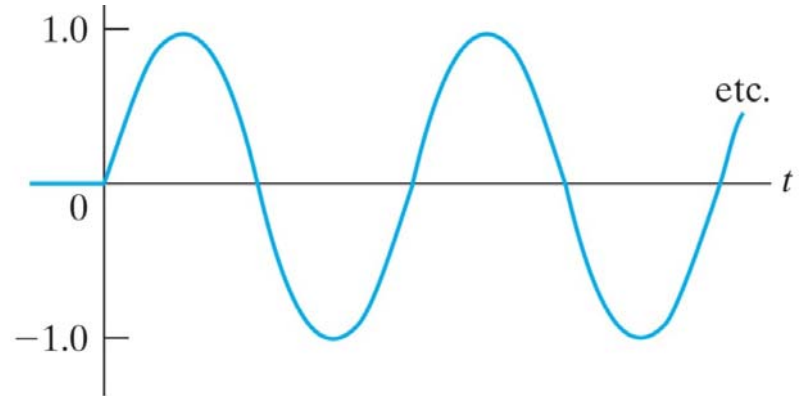
$$L\{\sin \omega t\} = \int_0^{\infty} (\sin \omega t) e^{-st} dt$$

$$= \int_0^{\infty} \frac{e^{j\omega t} - e^{-j\omega t}}{2j} e^{-st} dt$$

$$= \frac{1}{2j} \int_0^{\infty} \left[e^{-(s-j\omega)t} - e^{-(s+j\omega)t} \right] dt$$

$$= \frac{1}{2j} \left(\frac{1}{s-j\omega} - \frac{1}{s+j\omega} \right)$$

$$= \boxed{\frac{\omega}{s^2 + \omega^2}}, \text{ if } \operatorname{Re}\{s\} > 0.$$



LT of basic functions (1)

13

Type	$f(t)$	$F(s)$
unit - step	$u(t)$	$\frac{1}{s}$
ramp	$t \times u(t)$	$\frac{1}{s^2}$
polynomial	$t^n \times u(t)$	$\frac{n!}{s^{n+1}}$
exponential	$e^{-at} \times u(t)$	$\frac{1}{s+a}$

LT of basic functions (2)

14

Type	$f(t)$	$F(s)$
sine	$\sin \omega t \times u(t)$	$\frac{\omega}{s^2 + \omega^2}$
cosine	$\cos \omega t \times u(t)$	$\frac{s}{s^2 + \omega^2}$
damped ramp	$t e^{-at} \times u(t)$	$\frac{1}{(s + \underline{a})^2}$
damped sine	$\underline{e^{-at}} \sin \omega t \times u(t)$	$\frac{\omega}{(s + \underline{a})^2 + \omega^2}$

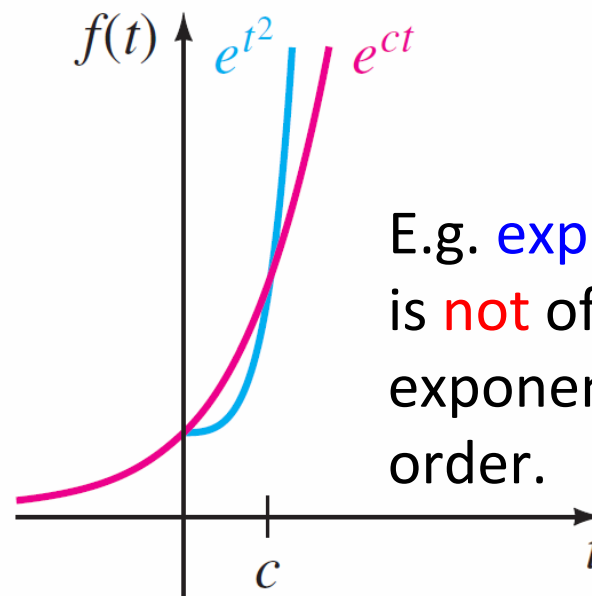
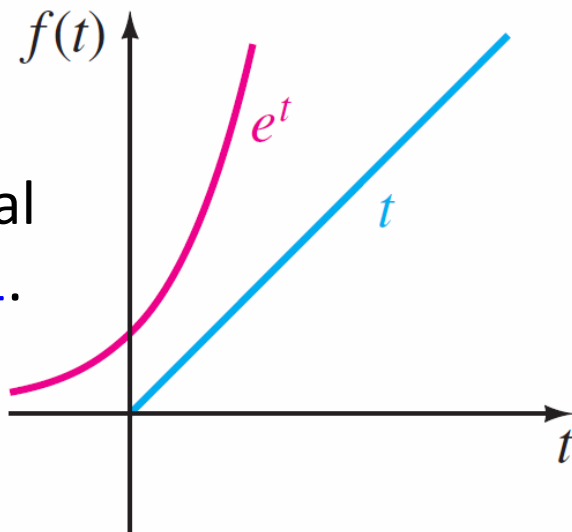
Exponential order

15

- A function $f(t)$ is of exponential order c if there are constants c , $M(>0)$, $T(>0)$ such that $f(t)$ is **eventually bounded by an exponential function**, i.e.

$$|f(t)| \leq Me^{ct}, \text{ for } t > T.$$

E.g. t is of exponential order $c = 1$.

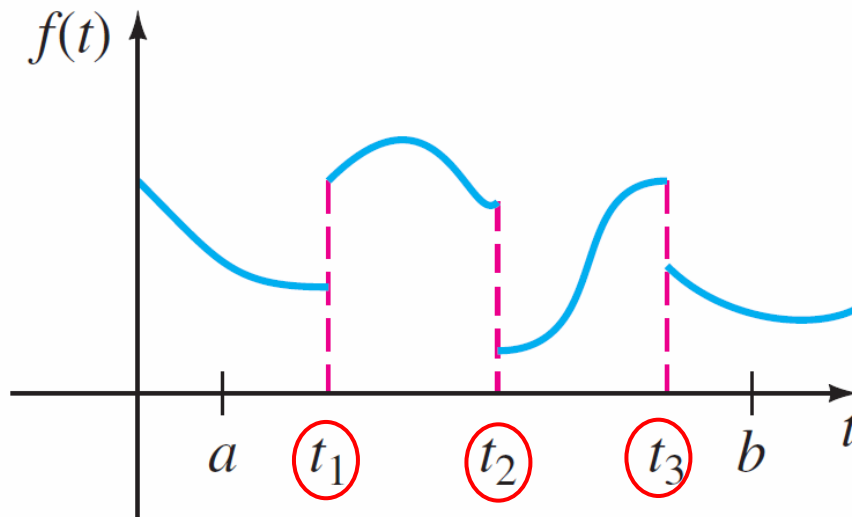


E.g. $\exp(t^2)$ is **not** of exponential order.

Existence of LT

16

- $L\{t^{-1}\}$ and $L\{t^t\}$ do not exist (not integrable).
- The **sufficient** (but not necessary) conditions that the $L\{f(t)\}$ exist for $\text{Re}\{s\} > c$ are:
 - (1) $f(t)$ is “piecewise continuous” for $t \in [0, \infty)$
 - (2) $f(t)$ is of “exponential order c ” for $t > T$.



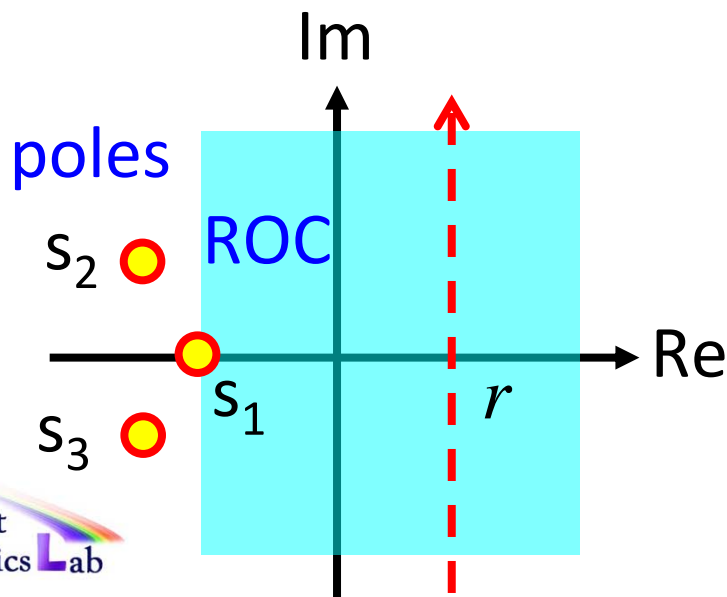
Piecewise continuous:
only finite number of
discontinuities $t = \{t_1, t_2, \dots, t_n\}$

□ Inverse LT

- Formula of complex integral
- Practical treatment

- Notation: $L^{-1}\{F(s)\} = f(t)$
- The formula involves with **complex integral** and **residue theorem** (to be studied in **EE 202**):

$$f(t) = L^{-1}\{F(s)\} = \frac{1}{2\pi i} \int_{r-i\infty}^{r+i\infty} F(s) \times e^{st} ds = \sum_{\text{all poles } s_k} \text{Res}[e^{st} \times F(s)]$$



- If $G(s)$ has a pole at s_i , \Rightarrow

$$G(s) = \sum_{n=1}^{\infty} a_n (s - s_i)^n + \sum_{n=1}^{\infty} \frac{b_n}{(s - s_i)^n},$$

$$\Rightarrow \text{Res}_{s_i}[G(s)] = b_1,$$

residue of $G(s)$ at s_i .

- Inverse LT is usually performed by using the table of LT of basic functions.
- Table (p13-4):
 - (1) $L^{-1}\{s^{-n}\} = t^{n-1}/n!$,
 - (2) $L^{-1}\{1/(s+a)\} = e^{-at}$,
 - (3) $L^{-1}\{1/(s^2+\omega^2)\} = \sin(\omega t)/\omega$,
 - (4) $L^{-1}\{s/(s^2+\omega^2)\} = \cos(\omega t)$, ...
- For $F(s)$ other than these basic forms, trying to simplify $F(s)$ by linearity and partial fractions.

Linearity of ILT

20

- $L^{-1}\{a \times F_1(s) + b \times F_2(s)\} = a \times L^{-1}\{F_1(s)\} + b \times L^{-1}\{F_2(s)\} = a \times f_1(t) + b \times f_2(t).$
- E.g. What's $f(t)$ if $F(s) = (-2s+6)/(s^2+4)$?

$$\begin{aligned} f(t) &= L^{-1}\left\{\frac{-2s+6}{s^2+4}\right\} \\ &= -2L^{-1}\left\{\frac{s}{s^2+2^2}\right\} + 3L^{-1}\left\{\frac{2}{s^2+2^2}\right\} \\ &= -2\cos(2t) + 3\sin(2t) \end{aligned}$$

Partial fractions, Example 3 (1)

21

$$F(s) = \frac{s^2 + 6s + 9}{(s^3 + s^2 - 10s + 8) = (s-1)(s-2)(s+4)}$$
$$= \frac{A}{s-1} + \frac{B}{s-2} + \frac{C}{s+4} = \frac{N(s)}{(s-1)(s-2)(s+4)}$$

$$N(s) = A(s-2)(s+4) + B(s-1)(s+4) + C(s-1)(s-2)$$
$$= s^2 + 6s + 9$$

- Method 1: Expanding $N(s)$ as a 2nd-order polynomial, comparing the coefficients, solving a system of 3 algebraic equations.

Example 3 (2)

- Method 2: Substituting $s = 1, 2, -4$ into $N(s)$.
- $N(1) = A(1-2)(1+4) = -5A = 1^2 + 6 \times 1 + 9 = 16$,
 $\Rightarrow A = -16/5$;
- $N(2) = B(2-1)(2+4) = 6B = 2^2 + 6 \times 2 + 9 = 25$,
 $\Rightarrow B = 25/6$;
- $N(-4) = C(-4-1)(-4-2) = 30C = (-4)^2 + 6 \times (-4) + 9 = 1$,
 $\Rightarrow C = 1/30$;

$$\begin{aligned}\Rightarrow f(t) &= L^{-1}\left\{\frac{A}{s-1}\right\} + L^{-1}\left\{\frac{B}{s-2}\right\} + L^{-1}\left\{\frac{C}{s+4}\right\} \\ &= -\frac{16}{5}e^t + \frac{25}{6}e^{2t} + \frac{1}{30}e^{-4t}\end{aligned}$$

□ Operational properties

- Differentiation
- Shift
- Convolution
- Integral
- Periodic functions

- By definition, $L\{f'(t)\} = \int_0^{\infty} f'(t) \times e^{-st} dt$;
- Integration by parts: $u = e^{-st}$, $v' = f'$, $\Rightarrow u' = -se^{-st}$, $v = f$:

$$\begin{aligned} \int_0^{\infty} f'(t) \times e^{-st} dt &= \left[f(t) \times e^{-st} \Big|_0^{\infty} \right] + s \left[\int_0^{\infty} f(t) \times e^{-st} dt \right] \\ &= [0 - f(0)] + sF(s) = sF(s) - f(0); \end{aligned}$$

if $\text{Re}\{s\} > 0$ and $F(s) \equiv L\{f(t)\}$.

- $L\{f'\} = sF(s) - f(0)$: **derivative** in the time domain is equivalent to **multiplication** in the s-domain.

- By definition, $L\{f''(t)\} = \int_0^{\infty} f''(t) \times e^{-st} dt$;
- Integration by parts: $u = e^{-st}$, $v' = f''$, $\Rightarrow u' = -se^{-st}$, $v = f'$:

$$\begin{aligned} \int_0^{\infty} f''(t) \times e^{-st} dt &= \left[f'(t) \times e^{-st} \Big|_0^{\infty} \right] + s \left[\int_0^{\infty} f'(t) \times e^{-st} dt \right] \\ &= [0 - f'(0)] + s[sF(s) - f(0)] = s^2 F(s) - sf(0) - f'(0); \end{aligned}$$

if $\text{Re}\{s\} > 0$.

- $L\{f''\} = s^2 F(s) - sf(0) - f'(0)$: derivative in the time domain is equivalent to multiplication (plus a polynomial) in the s-domain.

Example 5 (Sec. 7.2)

26

- ODE: $y'' - 3y' + 2y = e^{-4t}$; ICs: $\{y(0) = 1, y'(0) = 5\}$.
- Step 1: $L\{y'' - 3y' + 2y\} = [s^2Y(s) - sy(0) - y'(0)] - 3[sY(s) - y(0)] + 2Y(s) = (s^2 - 3s + 2)Y(s) - (s + 2) = L\{e^{-4t}\} = (s + 4)^{-1}$.

- Step 2: Solve $Y(s)$

$$(s^2 - 3s + 2)Y(s) = (s + 2) + \frac{1}{s + 4};$$

$$\Rightarrow Y(s) = \frac{s + 2}{s^2 - 3s + 2} + \frac{1}{(s^2 - 3s + 2)(s + 4)} = \frac{s^2 + 6s + 9}{(s - 1)(s - 2)(s + 4)}$$

- Step 3: $y(t) = L^{-1}\{Y(s)\} = -\frac{16}{5}e^t + \frac{25}{6}e^{2t} + \frac{1}{30}e^{-4t}$ (p21).

Advantages of LT approach

27

- Unique solution $y(t)$ is directly obtained without by way of $y_c(t)$ and $y_p(t)$.
- ICs have been taken into account by LT.
- $Y(s)$ is obtained by solving an algebraic equation.

- By definition,

$$\begin{aligned} L\{e^{at} \times f(t)\} &= \int_0^{\infty} e^{at} \times f(t) \times e^{-st} dt \\ &= \int_0^{\infty} f(t) \times e^{-(s-a)t} dt \\ &= F(s) \Big|_{s \rightarrow s-a} = F(s-a). \end{aligned}$$

- $L\{e^{at} \times f(t)\} = F(s-a)$: exponential modulation in the time domain is equivalent to translation (shift) in the s-domain.

Examples

29

$$\blacksquare L\{e^{at}\} = L\{e^{at} \times u(t)\} = L\{u(t)\}\Big|_{s \rightarrow s+a} = \frac{1}{s}\Big|_{s \rightarrow s+a} = \frac{1}{s+a},$$

same as that obtained by definition (p11).

$$\blacksquare L\{t \times e^{-at}\} = L\{t\}\Big|_{s \rightarrow s+a} = \frac{1}{s^2}\Big|_{s \rightarrow s+a} = \frac{1}{(s+a)^2},$$

$$\blacksquare L\{e^{-at} \times \sin \omega t\} = L\{\sin \omega t\}\Big|_{s \rightarrow s+a} = \frac{\omega}{s^2 + \omega^2}\Big|_{s \rightarrow s+a} \\ = \frac{\omega}{(s+a)^2 + \omega^2},$$

as shown in the last two rows of the LT table (p14).

Example 3 (Sec. 7.3) (1)

30

- ODE: $y'' - 6y' + 9y = t^2 e^{3t}$; ICs: $\{y(0) = 2, y'(0) = 17\}$.
- Step 1: $L\{y'' - 6y' + 9y\} = [s^2 Y(s) - 2s - 17] - 6[sY(s) - 2] + 9Y(s) = (s^2 - 6s + 9)Y(s) - (2s + 5) = L\{t^2 e^{3t}\} = 2/(s-3)^3$.
- Step 2: Solve $Y(s)$. $(s-3)^2 Y(s) = (2s+5) + 2/(s-3)^3, \Rightarrow$

$$Y(s) = \frac{2s+5}{(s-3)^2} + \frac{2}{(s-3)^5} = \left[\frac{A}{s-3} + \frac{B}{(s-3)^2} \right] + \frac{2}{(s-3)^5};$$

$$\left[\right] = \frac{A(s-3) + B}{(s-3)^2} = \frac{As + (B-3A)}{(s-3)^2} = \frac{2s+5}{(s-3)^2}, \Rightarrow \begin{cases} A = 2, \\ B = 11 \end{cases}$$

Example 3 (Sec. 7.3) (2)

31

■ Step 3: $y(t) = L^{-1}\{Y(s)\}$

$$\begin{aligned}y(t) &= L^{-1}\left\{\frac{2}{s-3}\right\} + L^{-1}\left\{\frac{11}{(s-3)^2}\right\} + L^{-1}\left\{\frac{2}{(s-3)^5}\right\} \\&= \left[2L^{-1}\{s^{-1}\} + 11L^{-1}\{s^{-2}\} + 2L^{-1}\{s^{-5}\}\right] \times e^{3t} \\&= \left[2u(t) + 11tu(t) + 2\frac{t^4 u(t)}{4!}\right] e^{3t} \\&= \left(2 + 11t + \frac{t^4}{12}\right) e^{3t} u(t)\end{aligned}$$

Example 4 (Sec. 7.3) (1)

32

- ODE: $y'' + 4y' + 6y = 1 + e^{-t}$; ICs: $\{y(0) = 0, y'(0) = 0\}$.
- Step 1: $L\{y'' + 4y' + 6y\} = [s^2Y(s) - 0s - 0] + 4[sY(s) - 0] + 6Y(s) = (s^2 + 4s + 6)Y(s) = L\{1 + e^{-t}\} = s^{-1} + (s+1)^{-1} = (2s+1)/[s(s+1)]$.
- Step 2: Solve $Y(s)$, $(s^2 + 4s + 6)Y(s) = (2s+1)/[s(s+1)]$,

$$Y(s) = \frac{2s+1}{s(s+1)(s^2+4s+6)} = \frac{A}{s} + \frac{B}{s+1} + \frac{Cs+D}{s^2+4s+6};$$

$$A(s+1)(s^2+4s+6) + Bs(s^2+4s+6) + (Cs+D)s(s+1)$$

$$= 2s+1, \Rightarrow \left\{ A = \frac{1}{6}, B = \frac{1}{3}, C = -\frac{1}{2}, D = -\frac{5}{3} \right\}$$

Example 3 (Sec. 7.3) (2)

33

- Since $s^2 + 4s + 6 = 0$ has no real root, regrouping it as $(s - \alpha)^2 + \omega^2 = (s+2)^2 + (\sqrt{2})^2$,

$$F(s) \equiv \frac{Cs + D}{s^2 + 4s + 6} = \frac{C(s + 2) + (D - 2C)}{(s + 2)^2 + (\sqrt{2})^2}$$

- Step 3: $y(t) = L^{-1}\{Y(s)\} = L^{-1}\{A/s\} + L^{-1}\{B/(s+1)\} + f(t)$,

$$\begin{aligned} f(t) &= L^{-1}\left\{\frac{C(s + 2) + (D - 2C)}{(s + 2)^2 + (\sqrt{2})^2}\right\} = L^{-1}\left\{\frac{Cs + (D - 2C)}{s^2 + (\sqrt{2})^2}\right\}e^{-2t} \\ &= \left[C \times L^{-1}\left\{\frac{s}{s^2 + (\sqrt{2})^2}\right\} + (D - 2C) \times L^{-1}\left\{\frac{1}{s^2 + (\sqrt{2})^2}\right\} \right]e^{-2t}; \end{aligned}$$

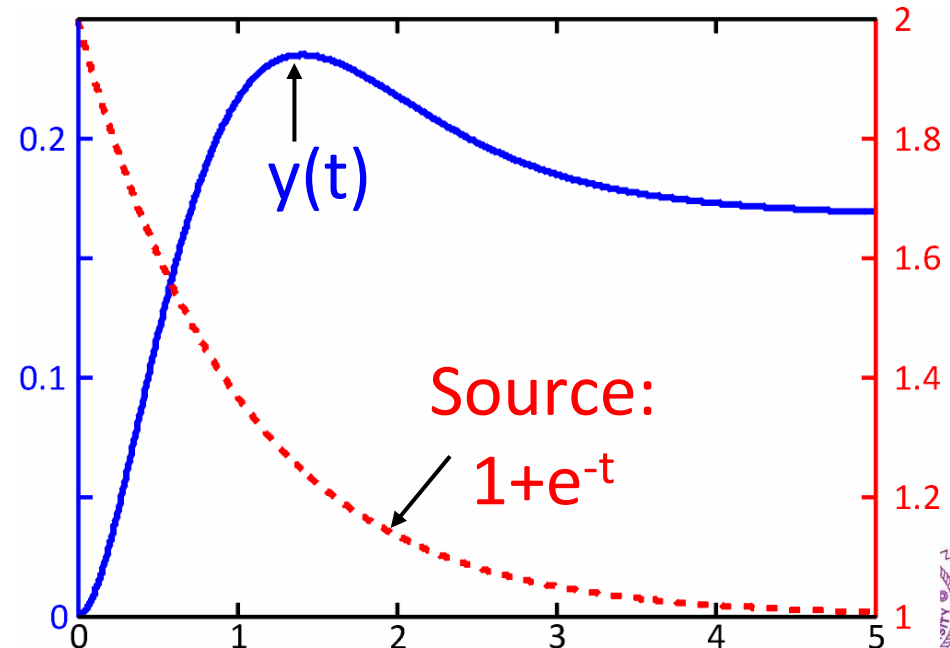
Example 3 (Sec. 7.3) (3)

34

$$f(t) = \left[C \times \cos(\sqrt{2}t) + (D - 2C) \times \frac{\sin(\sqrt{2}t)}{\sqrt{2}} \right] e^{-2t} \bigg|_{C=-\frac{1}{2}, D=-\frac{5}{3}}$$
$$= - \left[\frac{1}{2} \cos(\sqrt{2}t) + \frac{\sqrt{2}}{3} \sin(\sqrt{2}t) \right] e^{-2t}$$

$$\Rightarrow y(t) = \frac{1}{6} + \frac{1}{3} e^{-t}$$

$$- \left[\frac{1}{2} \cos(\sqrt{2}t) + \frac{\sqrt{2}}{3} \sin(\sqrt{2}t) \right] e^{-2t}$$



$L\{f(t-a) \times u(t-a)\}$

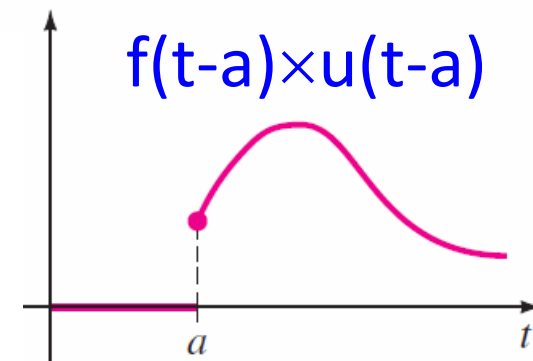
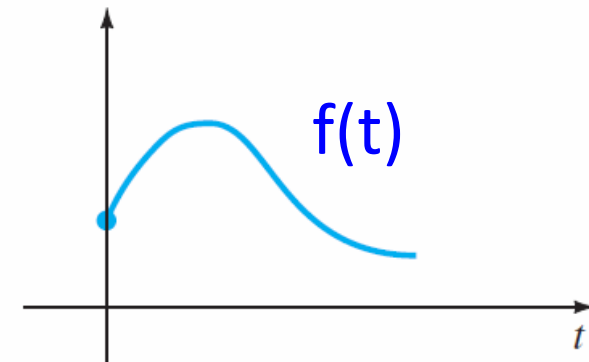
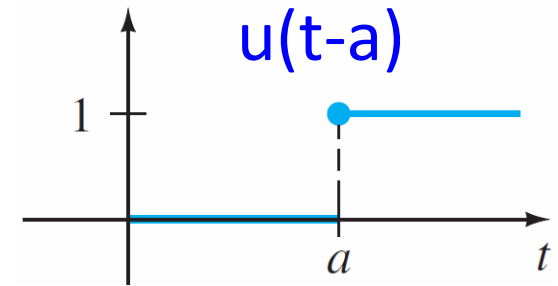
- By definition,

$$\begin{aligned} g(t) &\equiv L\{f(t-a) \times u(t-a)\} \\ &= \int_a^\infty f(t-a) \times e^{-st} dt; \end{aligned}$$

Let $t-a = v$, $\Rightarrow dt = dv$;

$$\begin{aligned} g(t) &= \int_0^\infty f(v) \times e^{-s(v+a)} dv \\ &= e^{-sa} \left[\int_0^\infty f(v) \times e^{-sv} dv \right] = e^{-as} \times F(s). \end{aligned}$$

- Shift in the t-domain, \rightarrow modulation in the s-domain. Complementary relation of $L\{e^{at} \times f(t)\} = F(s-a)$:



$L\{f(t) \times u(t-a)\}$

- By definition, $g(t) \equiv L\{f(t) \times u(t-a)\} = \int_a^\infty f(t) \times e^{-st} dt$;

Let $t-a = v, \Rightarrow dt = dv$;

$$\begin{aligned} g(t) &= \int_0^\infty f(v+a) \times e^{-s(v+a)} dv \\ &= e^{-sa} \left[\int_0^\infty f(v+a) \times e^{-sv} dv \right] = e^{-as} \times L\{f(t+a)\}. \end{aligned}$$

- Activating $f(t)$ at “ $t = a$ ” is equivalent to activating $f(t+a)$ at $t = 0$.

Example 8 (Sec. 7.3) (1)

37

- ODE: $y' + y = 3\cos(t) \times u(t-\pi)$; IC: $y(0) = 5$.
- Step 1: $L\{y' + y\} = [sY(s) - 5] + Y(s) = (s+1)Y(s) - 5 = L\{3\cos(t) \times u(t-\pi)\} = 3e^{-\pi s} \times L\{\cos(t+\pi) = -\cos(1t)\} = -3(e^{-\pi s}) \times [s/(s^2+1)]$.
- Step 2: Solve $Y(s)$, $(s+1)Y(s) = 5 - 3e^{-\pi s}s/(s^2+1)$,

$$Y(s) = \frac{5}{s+1} - F(s) \times e^{-\pi s}, \text{ where } F(s) = \frac{3s}{(s+1)(s^2+1)}$$

$$= \frac{A}{s+1} + \frac{Bs+C}{s^2+1}, \Rightarrow A(s^2+1) + (Bs+C)(s+1) = 3s,$$

$$\Rightarrow \left\{ A = -\frac{3}{2}, B = \frac{3}{2}, C = \frac{3}{2} \right\}$$

Example 8 (Sec. 7.3) (2)

38

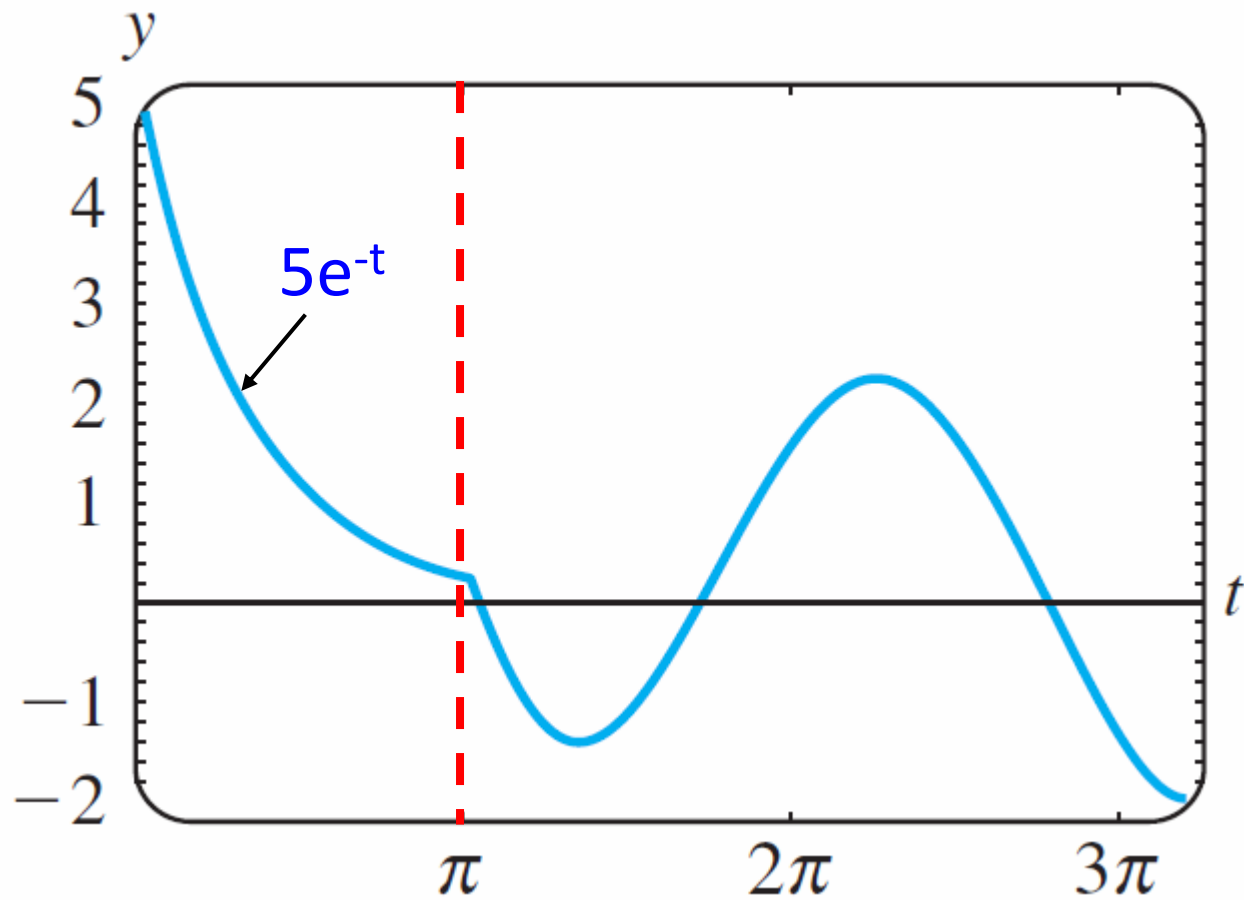
$$\Rightarrow Y(s) = \frac{5}{s+1} + \frac{3}{2} e^{-\pi s} \left(\frac{1}{s+1} - \frac{1}{s^2+1} - \frac{s}{s^2+1} \right).$$

■ Step 3: $y(t) = \mathcal{L}^{-1}\{Y(s)\}$

$$\begin{aligned} y(t) &= \mathcal{L}^{-1} \left\{ \frac{5}{s+1} \right\} + \frac{3}{2} \left[\mathcal{L}^{-1} \left\{ \frac{1}{s+1} - \frac{1}{s^2+1} - \frac{s}{s^2+1} \right\} \right]_{t \rightarrow t-\pi} \\ &= 5e^{-t} + \frac{3}{2} \left[e^{-t} - \sin(t) - \cos(t) \right] \Big|_{t \rightarrow t-\pi} \\ &= 5e^{-t} + \frac{3}{2} \left[e^{-(t-\pi)} - \sin(t-\pi) - \cos(t-\pi) \right] \times u(t-\pi) \\ &= 5e^{-t} + \frac{3}{2} \left[e^{-(t-\pi)} + \sin(t) + \cos(t) \right] \times u(t-\pi) \end{aligned}$$

Solution plot

39



- By definition, $\frac{d}{ds} F(s) = \frac{d}{ds} \int_0^\infty f(t) \times e^{-st} dt$;
- If the order of differentiation and integration is interchangeable,

$$\begin{aligned} F'(s) &= \int_0^\infty \frac{\partial}{\partial s} [f(t) \times e^{-st}] dt = \int_0^\infty f(t) \times \left(\frac{\partial}{\partial s} e^{-st} \right) dt \\ &= \int_0^\infty f(t) \times (-te^{-st}) dt = \int_0^\infty -tf(t) \times e^{-st} dt = L^{-1}\{-tf(t)\} \end{aligned}$$

- $L^{-1}\{F'\} = -t \times f(t)$: **derivative** in the s-domain is equivalent to **multiplication** in the time domain.

- Without using the inverse LT formula, let's guess $L^{-1}\{F''(s)\} = t^2 f(t)$ given $L^{-1}\{F'(s)\} = -t \times f(t)$ is true.

- Proof:

$$\begin{aligned} L\{t^2 f(t)\} &= \int_0^\infty t^2 f(t) \times e^{-st} dt = \int_0^\infty t \times g(t) \times e^{-st} dt \\ &= L\{t \times g(t)\} = -G'(s), \text{ where } g(t) \equiv t \times f(t). \end{aligned}$$

Since $G(s) = -F'(s)$, $\Rightarrow L\{t^2 f(t)\} = F''(s)$.

- $L^{-1}\{F''\} = t^2 f(t)$: **derivative** in the s-domain is equivalent to **multiplication** in the time domain.

Example: $L\{t^2\}$

42

- Previously, we got $L\{t^2\} = 2/s^3$ by definition (p10).
Here is a faster way:
- Let $f(t) = 1, \Rightarrow F(s) = 1/s$.
- $g(t) \equiv t^2 = t^2 \times f(t), \Rightarrow G(s) = F''(s) = (s^{-1})'' = -(s^{-2}) = 2s^{-3}$.
- For $h(t) \equiv t^n = t^n \times f(t), \Rightarrow$

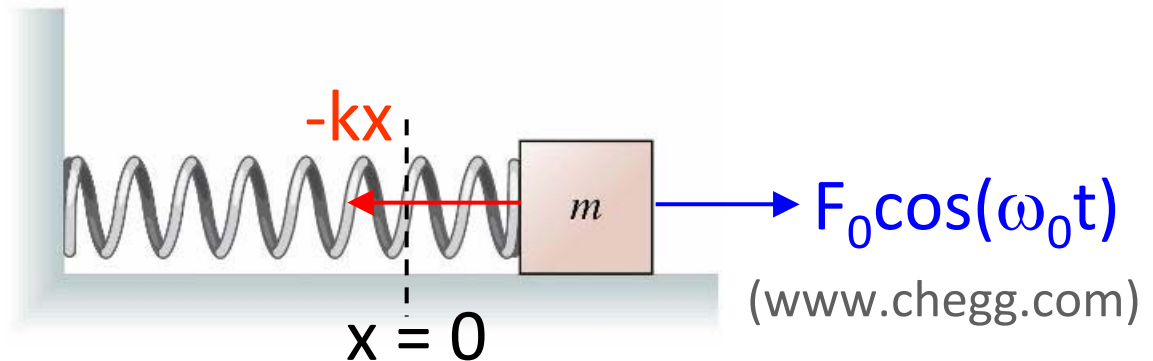
$$\begin{aligned} H(s) &= (-1)^n \frac{d^n}{ds^n} F(s) = (-1)^n \frac{d^n}{ds^n} s^{-1} \\ &= (-1)^{n+1} \frac{d^{n-1}}{ds^{n-1}} s^{-2} = (-1)^{n+2} (2) \frac{d^{n-2}}{ds^{n-2}} s^{-3} = \dots \\ &= (-1)^{n+n} n! s^{-(n+1)} = n! s^{-(n+1)}. \end{aligned}$$

Resonantly driven mass-spring (1)

43

- ODE: $x''(t) + (\omega_0)^2 x(t) = A \times \cos(\omega_0 t)$, where $\omega_0 = \sqrt{k/m}$,
 $A = F_0/m$.

- ICs: $x(0) = X_0$,
 $x'(0) = V_0$.



- Step 1: $L\{\text{ODE}\}$

$$\begin{aligned} L\{x'' + \omega_0^2 x\} &= (s^2 X - \underbrace{X_0}_{\text{blue dashed circle}} s - \underbrace{V_0}_{\text{green dashed circle}}) + \omega_0^2 X \\ &= (s^2 + \omega_0^2) X - (X_0 s + V_0); \end{aligned}$$

$$L\{A \times \cos \omega_0 t\} = \frac{As}{s^2 + \omega_0^2}$$

Resonantly driven mass-spring (2)

44

■ Step 2: Solve $X(s)$

$$(s^2 + \omega_0^2)X - (X_0s + V_0) = \frac{As}{s^2 + \omega_0^2}$$

$$\Rightarrow X(s) = \frac{X_0s + V_0}{s^2 + \omega_0^2} + \frac{As}{(s^2 + \omega_0^2)^2} \equiv X_1(s) + X_2(s)$$

■ Step 3: $x(t) = L^{-1}\{X(s)\}$

$$x_1(t) = L^{-1}\left\{\frac{X_0s + V_0}{s^2 + \omega_0^2}\right\} = X_0 \cos \omega_0 t + \frac{V_0}{\omega_0} \sin \omega_0 t$$

- Initial position X_0 and initial velocity V_0 contribute to $\cos(\omega_0 t)$ and $\sin(\omega_0 t)$ terms, respectively.

Resonantly driven mass-spring (3)

45

- To calculate $L^{-1}\{X_2(s)\}$, let's look for $G(s) \equiv \int X_2(s)ds$ such that $x_2(t) = -t \times g(t)$.

$$G(s) = \int \frac{As}{(s^2 + \omega_0^2)^2} ds = -\frac{A}{2} \frac{1}{s^2 + \omega_0^2};$$

$$\Rightarrow g(t) = -\frac{A}{2\omega_0} L^{-1}\left\{\frac{\omega_0}{s^2 + \omega_0^2}\right\} = -\frac{A}{2\omega_0} \sin \omega_0 t;$$

$$\Rightarrow x_2(t) = \frac{A}{2\omega_0} t \sin \omega_0 t;$$

- Resonant driving force causes linearly growing oscillation, **unavailable by using Ch4's approach.**

Example 2 (Sec. 7.4)

46

- ODE: $x''(t) + 16x(t) = \cos(4t)$; $\Rightarrow \omega_0 = 4, A = 1$.
- ICs: $x(0) = X_0 = 0, x'(0) = V_0 = 1$.
- By the general analysis, $x(t) = x_1(t) + x_2(t)$

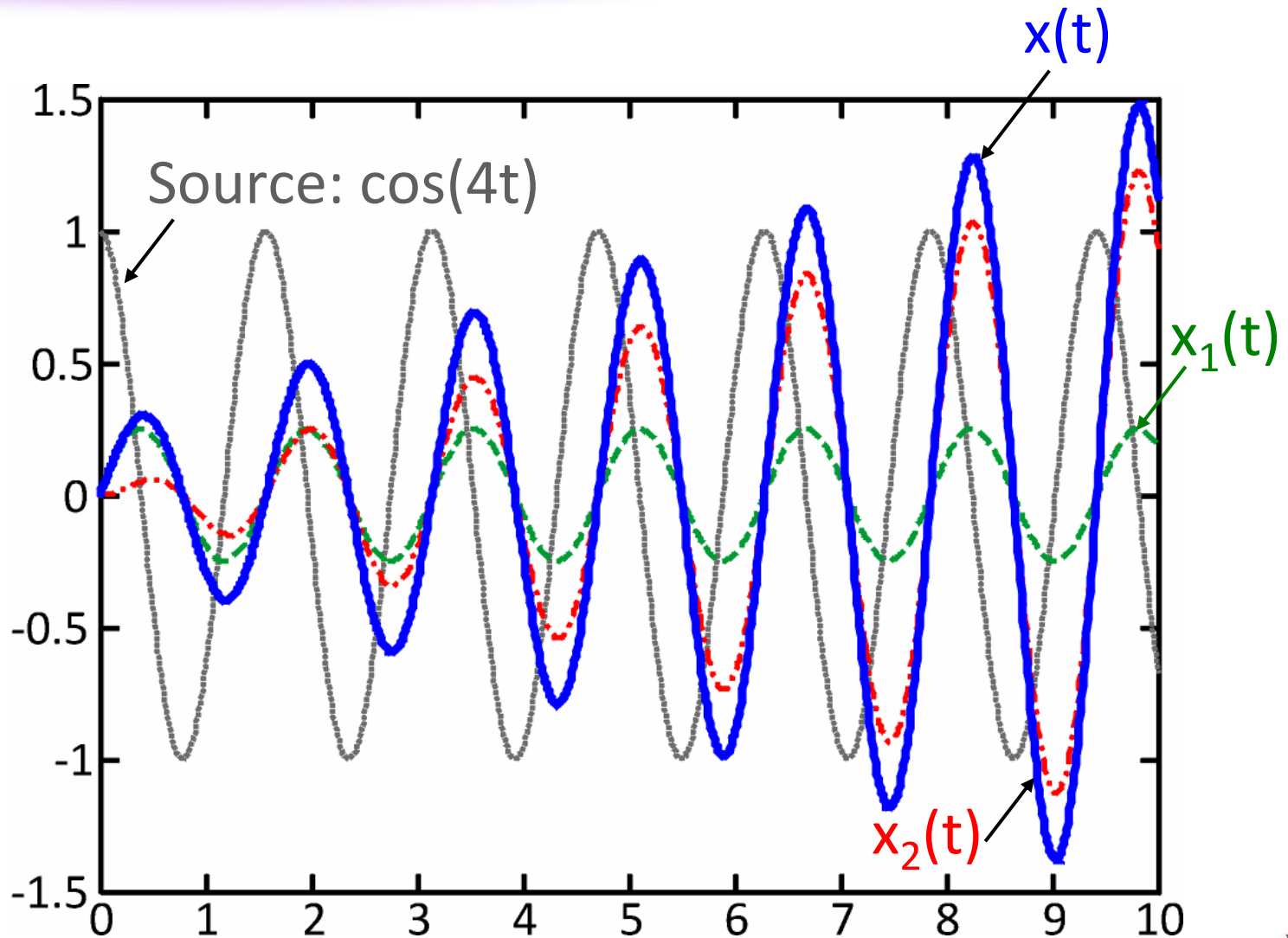
$$= \underbrace{X_0}_{\text{blue dotted}} \cos \omega_0 t + \underbrace{\frac{V_0}{\omega_0}}_{\text{green dotted}} \sin \omega_0 t + \frac{A}{2\omega_0} t \sin \omega_0 t$$

$$= 0 \times \cos 4t + \frac{1}{4} \sin 4t + \frac{1}{2 \times 4} t \sin 4t$$

$$= \frac{1}{4} \sin 4t + \frac{1}{8} t \sin 4t$$

Solution plot

47



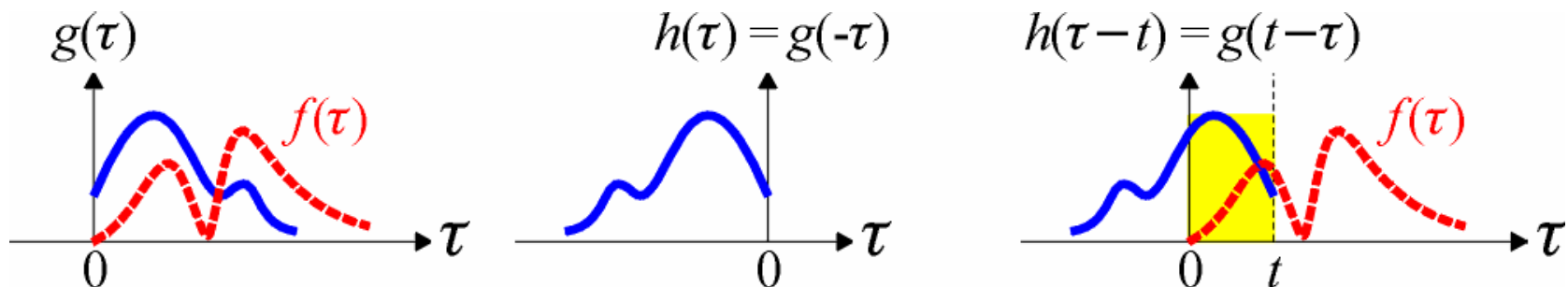
Convolution $f * g$

48

- If $f(\tau)$ and $g(\tau)$ are piecewise continuous over $\tau = [0, \infty)$, the convolution of f and g is defined as

$$f * g = \int_0^t f(\tau) \times g(t - \tau) d\tau = \int_0^t f(t - \tau) \times g(\tau) d\tau$$

- Why is the integral range $\tau = (0, t)$?

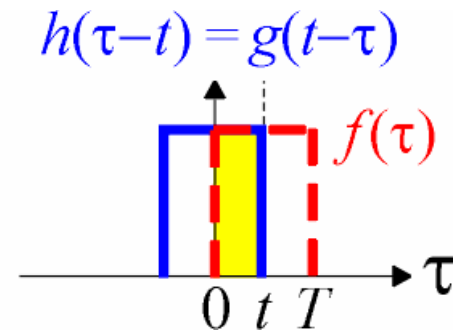
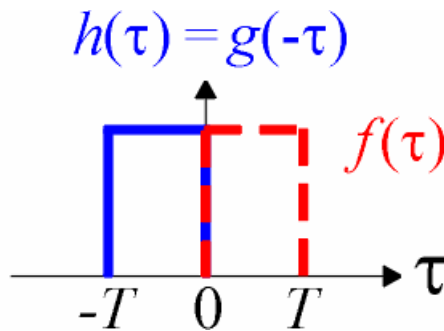
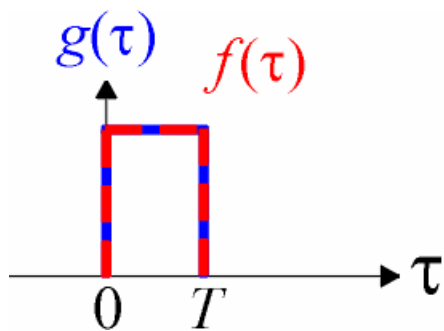


Example: $\Pi * \Pi$

49

- Let $\Pi(t) = \{1, \text{ for } 0 < t < T; 0, \text{ otherwise}\}$ is a rectangular function of width T .

$$\Pi * \Pi = \int_0^t \Pi(\tau) \times \Pi(t - \tau) d\tau$$



$$L\{f * g\} = F \times G \quad (1)$$

■ Proof:

$$F(s) \equiv L\{f(t)\} = \int_0^{\infty} f(\tau) \times e^{-s\tau} d\tau,$$

$$G(s) \equiv L\{g(t)\} = \int_0^{\infty} g(\beta) \times e^{-s\beta} d\beta.$$

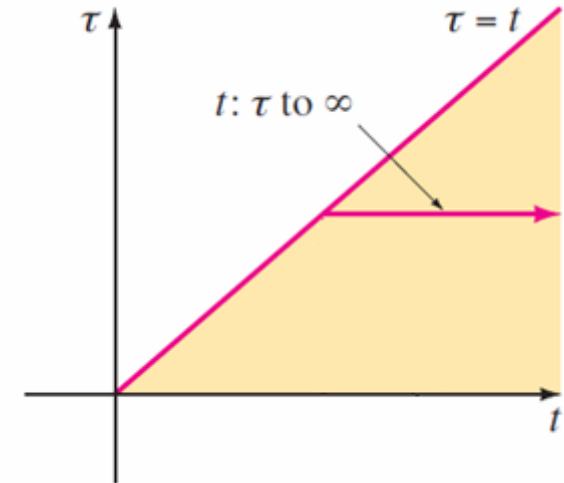
$$\begin{aligned} \Rightarrow F(s) \times G(s) &= \left[\int_0^{\infty} f(\tau) \times e^{-s\tau} d\tau \right] \times \left[\int_0^{\infty} g(\beta) \times e^{-s\beta} d\beta \right] \\ &= \int_0^{\infty} \int_0^{\infty} f(\tau) g(\beta) \times e^{-s(\tau+\beta)} d\tau d\beta \\ &= \int_0^{\infty} f(\tau) \times \left[\int_0^{\infty} g(\beta) \times e^{-s(\tau+\beta)} d\beta \right] d\tau \end{aligned}$$

$$L\{f * g\} = F \times G \quad (2)$$

- Let $t = \tau + \beta, \Rightarrow d\beta = dt,$

$$F(s) \times G(s)$$

$$= \int_0^\infty f(\tau) \times \left[\int_\tau^\infty g(t - \tau) \times e^{-st} dt \right] d\tau$$

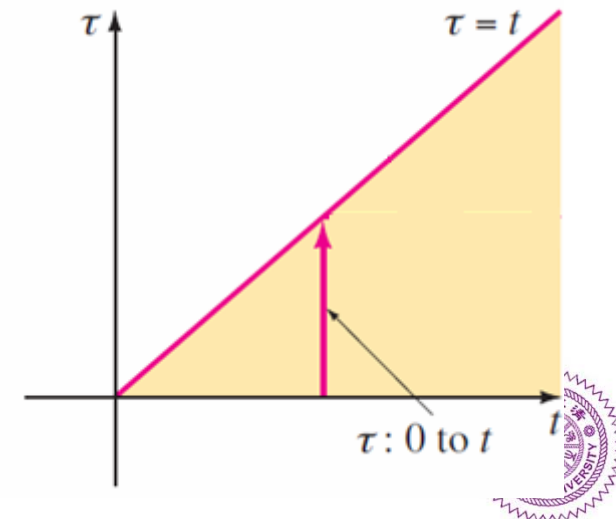


- Interchanging the integration order:

$$F(s) \times G(s)$$

$$= \int_0^\infty \left[\int_0^t f(\tau) \times g(t - \tau) d\tau \right] \times e^{-st} dt$$

$$= \int_0^\infty (f * g) \times e^{-st} dt = L\{f * g\}.$$

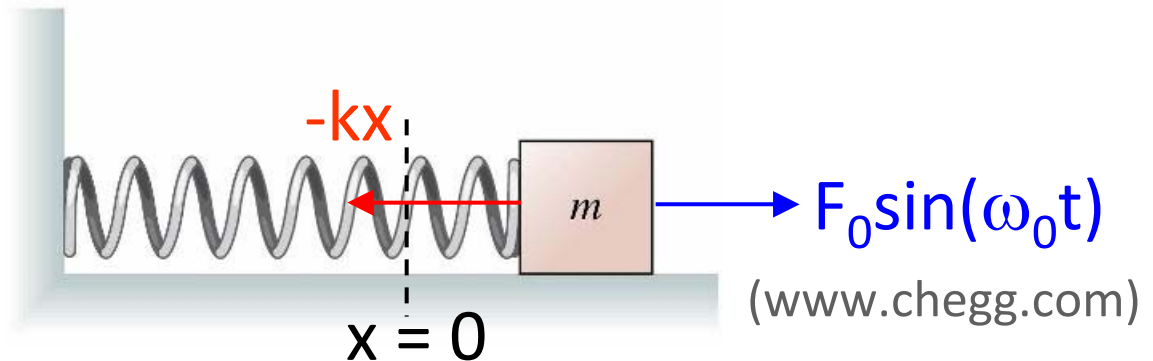


Revisit: Resonantly driven mass-spring (1)

52

- ODE: $x''(t) + (\omega_0)^2 x(t) = A \times \sin(\omega_0 t)$, where $\omega_0 = \sqrt{k/m}$,
 $A = F_0/m$.

- ICs: $x(0) = X_0$,
 $x'(0) = V_0$.



- Step 1: $L\{\text{ODE}\}$

$$L\{x'' + \omega_0^2 x\} = (s^2 + \omega_0^2)X - (X_0 s + V_0); \dots \text{same}$$

$$L\{A \times \sin \omega_0 t\} = \frac{A \omega_0}{s^2 + \omega_0^2} \text{ v.s. } L\{A \times \cos \omega_0 t\} = \frac{As}{s^2 + \omega_0^2}$$

- Step 2: Solve $X(s)$

$$(s^2 + \omega_0^2)X - (X_0s + V_0) = \frac{A\omega_0}{s^2 + \omega_0^2}$$

$$\Rightarrow X(s) = \frac{X_0s + V_0}{s^2 + \omega_0^2} + \frac{A\omega_0}{(s^2 + \omega_0^2)^2} \equiv X_1(s) + X_2(s)$$

- Step 3: $x(t) = \mathcal{L}^{-1}\{X(s)\} = x_1(t) + x_2(t)$, where $x_1(t) = X_0\cos(\omega_0t) + (V_0/\omega_0)\sin(\omega_0t)$ is the same as before.
- However, no easy anti-derivative $G(s) \equiv \int X_2(s)ds$ exists, and $x_2(t) = -t \times g(t)$ does not apply.

Revisit: Resonantly driven mass-spring (3)

54

■ Instead, $X_2(s) = \frac{A\omega_0}{(s^2 + \omega_0^2)^2} \equiv A\omega_0 \times F(s) \times G(s),$

where $F(s) = G(s) = 1/[s^2 + (\omega_0)^2],$

$\Rightarrow f(t) = g(t) = \sin(\omega_0 t)/\omega_0; \quad x_2(t) = A\omega_0(f * g) = (A/\omega_0)I,$

$I \equiv \sin(\omega_0 t) * \sin(\omega_0 t) = \int_0^t \sin(\omega_0 \tau) \times \sin[\omega_0(t - \tau)] d\tau;$

■ By the “product to sum” formula:

$$\sin \alpha \times \sin \beta = \frac{\cos(\alpha - \beta) - \cos(\alpha + \beta)}{2};$$

the integrand is $\frac{\cos(2\omega_0 \tau - \omega_0 t) - \cos(\omega_0 t)}{2};$

Revisit: Resonantly driven mass-spring (4)

55

$$\blacksquare \quad I = \frac{1}{2} \int_0^t [\cos(2\omega_0 \tau - \omega_0 t) - \cos(\omega_0 t)] d\tau$$

variable

$$= \frac{1}{2} \left\{ \left[\frac{\sin(2\omega_0 \tau - \omega_0 t)}{2\omega_0} \right]_0^t - \cos(\omega_0 t) \times \left(\tau \Big|_0^t \right) \right\}$$

$$= \frac{1}{2} \left[\frac{\sin(\omega_0 t) + \sin(\omega_0 t)}{2\omega_0} - t \times \cos(\omega_0 t) \right];$$

$$\Rightarrow x_2(t) = \frac{A}{2\omega_0} \left[\frac{\sin(\omega_0 t)}{\omega_0} - t \times \cos(\omega_0 t) \right] \text{ v.s. } \frac{A}{2\omega_0} t \sin \omega_0 t$$

- Let $g(t) = 1, \Rightarrow G(s) = 1/s$;

$$\Rightarrow f * g = \int_0^t [f(\tau) \times g(t - \tau)] d\tau = \int_0^t f(\tau) d\tau.$$

- By $L\{f * g\} = F \times G, \Rightarrow L\left\{\int_0^t f(\tau) d\tau\right\} = \frac{F(s)}{s}.$

- $L\{\int f\} = F(s)/s$: **integral** in the time domain is equivalent to **division** in the s-domain.

- It's the complementary relation of $L\{f'\} = sF(s) - f(0).$

Example: $L^{-1}\{F(s)/s\}$

57

- $G(s) \equiv \frac{1}{s(s^2 + 1)}, \Rightarrow g(t) \equiv L^{-1}\{G(s)\} = ?$

- Let $F(s) \equiv 1/(s^2+1), \Rightarrow f(t) = \sin(t);$

- $G(s) = F(s)/s,$

$$\Rightarrow g(t) = \int_0^t f(\tau) d\tau = \int_0^t \sin(\tau) d\tau = -\cos(\tau) \Big|_0^t = 1 - \cos(t).$$

- Check: $G(s) = \frac{A}{s} + \frac{Bs + C}{s^2 + 1} = \frac{(A + B)s^2 + Cs + A}{s(s^2 + 1)} = 1;$

$$\Rightarrow \{A = 1, B = -1, C = 0\};$$

$$\Rightarrow g(t) = L^{-1}\left\{\frac{1}{s}\right\} - L^{-1}\left\{\frac{s}{s^2 + 1}\right\} = 1 - \cos(t). \text{ (same)}$$

Example: $L^{-1}\{F(s)/s^2\}$

58

- $H(s) \equiv \frac{1}{s^2(s^2 + 1)}, \Rightarrow h(t) \equiv L^{-1}\{H(s)\} = ?$

- Let $G(s) \equiv 1/[s(s^2+1)], \Rightarrow g(t) = 1-\cos(t);$

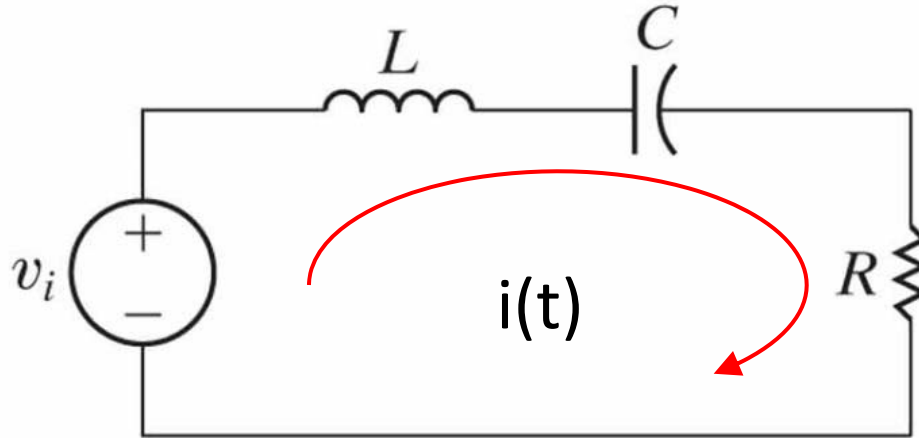
- $H(s) = G(s)/s,$

$$\begin{aligned} \Rightarrow h(t) &= \int_0^t g(\tau) d\tau = \int_0^t [1 - \cos(\tau)] d\tau \\ &= t - \left[\sin(\tau) \Big|_0^t \right] = t - \sin(t). \end{aligned}$$

Integrodifferential equations

59

- E.g. Series RLC circuit.



- By Kirchhoff voltage law, $v_L(t) = L \times i'(t)$, $i(t) = C \times v_C'(t)$:

$$Li'(t) + Ri(t) + \frac{1}{C} \int_0^t i(\tau) d\tau = v_i(t)$$

Example 6 (1)

- Let $v_i(t) = V_0 t[1-u(t-1)]$ (curve?), $i(0) = I_0$.
- Step 1:
$$L\left\{Li'(t) + Ri(t) + \frac{1}{C} \int_0^t i(\tau) d\tau\right\}$$
$$= L[sI(s) - I_0] + R \times I(s) + \frac{1}{C} \times \frac{I(s)}{s};$$

By $L\{f(t) \times u(t-a)\} = e^{-as} \times L\{f(t+a) \times u(t)\},$

$$L\{v_i(t)\} = V_0[L\{t\} - L\{t \times u(t-1)\}]$$
$$= V_0\left[\frac{1}{s^2} - e^{-s} \times L\{(t+1) \times u(t)\}\right] = V_0\left(\frac{1}{s^2} - \frac{e^{-s}}{s^2} - \frac{e^{-s}}{s}\right);$$

Example 6 (2)

61

■ Step 2: Solve $I(s)$

$$\left(Ls + R + \frac{1}{Cs} \right) I(s) - LI_0 = V_0 \left(\frac{1}{s^2} - \frac{e^{-s}}{s^2} - \frac{e^{-s}}{s} \right)$$

$$\Rightarrow L \left[\left(s^2 + \frac{R}{L}s + \frac{1}{LC} \right) I(s) - I_0 s \right] = V_0 \left(\frac{1}{s} - \frac{e^{-s}}{s} - e^{-s} \right)$$

$$\Rightarrow I(s) = I_0 \frac{s}{D(s)} + I_m \frac{(s^{-1} - s^{-1}e^{-s} - e^{-s})}{D(s)},$$

$$\text{where } I_m = \frac{V_0}{L}, \quad D(s) = s^2 + \frac{R}{L}s + \frac{1}{LC}.$$

Example 6 (3)

62

- Step 3: $L^{-1}\{I(s)\}$.
- Let $L = 0.1 \text{ H}$, $C = 0.1 \text{ F}$, $R = 2 \Omega$, $V_0 = 120$, $I_0 = 0$, \Rightarrow
 $I_m = 1200 \text{ A}$, $D(s) = s^2 + 20s + 100 = (s+10)^2$... **Case 2.**

$$I(s) = \frac{\cancel{I_0}s}{(s+10)^2} + I_m \frac{(s^{-1} - s^{-1}e^{-s} - e^{-s})}{(s+10)^2} = I_m (I_1 - I_2 - I_3),$$

Partial fractions

$$\begin{aligned} i_1(t) &= L^{-1} \left\{ \frac{1}{s(s+10)^2} \right\} = L^{-1} \left\{ \frac{0.01}{s} - \frac{0.01}{s+10} - \frac{0.1}{(s+10)^2} \right\} \\ &= [0.01 - 0.01e^{-10t} - 0.1te^{-10t}] \times u(t); \end{aligned}$$

Example 6 (4)

63

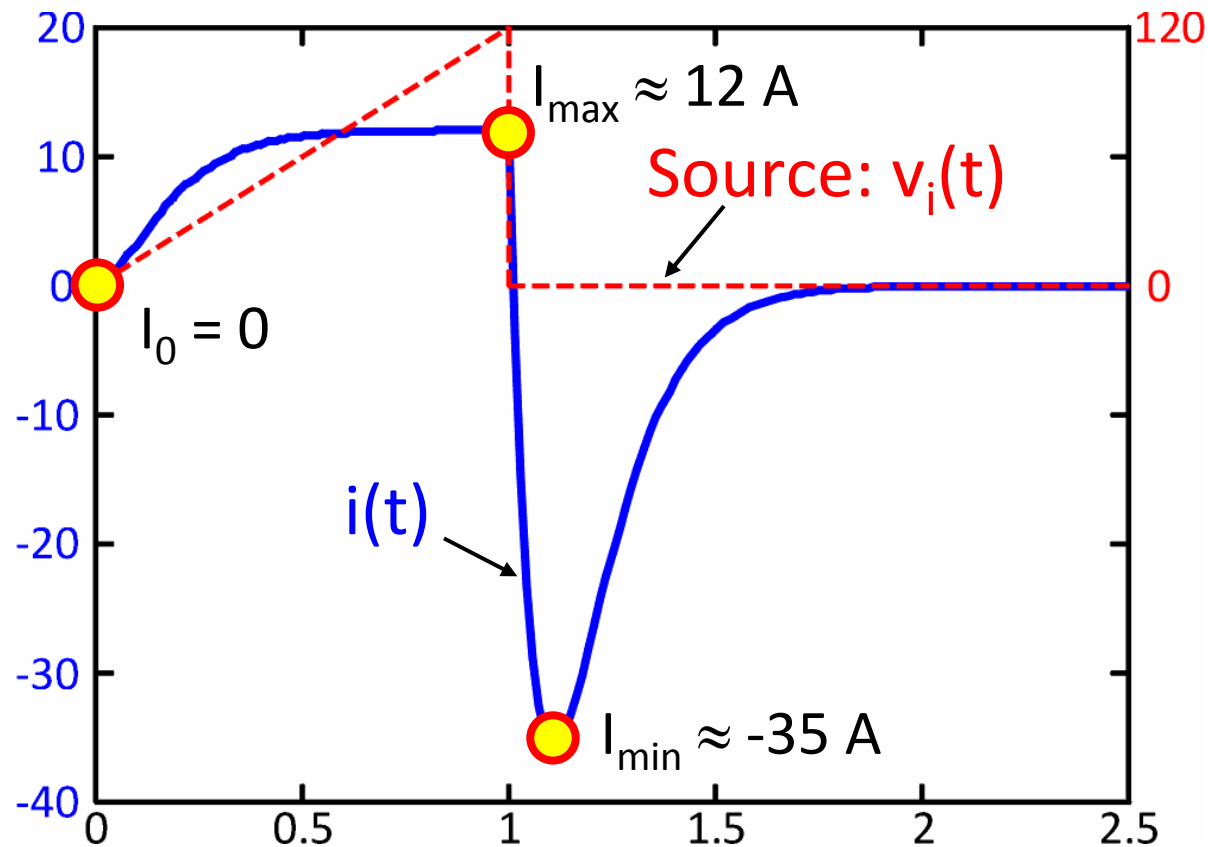
$$i_2(t) = L^{-1} \left\{ \frac{e^{-s}}{s(s+10)^2} \right\} = L^{-1} \{ e^{-s} \times I_1(s) \} = i_1(t-1);$$

$$i_3(t) = L^{-1} \left\{ \frac{e^{-s}}{(s+10)^2} \right\} = L^{-1} \left\{ \frac{1}{(s+10)^2} \right\} \Big|_{t \rightarrow t-1} = (t-1)e^{-10(t-1)};$$

$$\begin{aligned} i(t) &= I_m \{ i_1(t) - [i_2(t) + i_3(t)] \} \\ &= 1200 \left\{ (0.01 - 0.01e^{-10t} - 0.1te^{-10t}) \times u(t) \right. \\ &\quad \left. - [0.01 - 0.01e^{-10(t-1)} + 0.9(t-1)e^{-10(t-1)}] \times u(t-1) \right\}, \end{aligned}$$

Solution plot

$$i(t) = \begin{cases} 12[1 - e^{-10t} - 10te^{-10t}], & \text{if } 0 < t < 1. \\ 12[-e^{-10t} + e^{-10(t-1)} - 10te^{-10t} - 90(t-1)e^{-10(t-1)}], & \text{if } t > 1. \end{cases}$$



$L\{f(t)\}$, with $f(t+T) = f(t)$

- $F(s) = L\{f(t)\} = \frac{1}{1 - e^{-sT}} \int_0^T f(t) \times e^{-st} dt.$

- Proof:

$$F(s) = \int_0^T f(t) \times e^{-st} dt + \int_T^\infty f(t) \times e^{-st} dt \equiv I_1 + I_2.$$

Let $u = t - T, \Rightarrow dt = du$: $I_2 = \int_0^\infty \underline{f(u+T)} \times e^{-s(u+T)} du$

$$= e^{-sT} \left[\int_0^\infty \underline{f(u)} \times e^{-su} du \right] = e^{-sT} \times F(s) = e^{-sT} (I_1 + I_2);$$

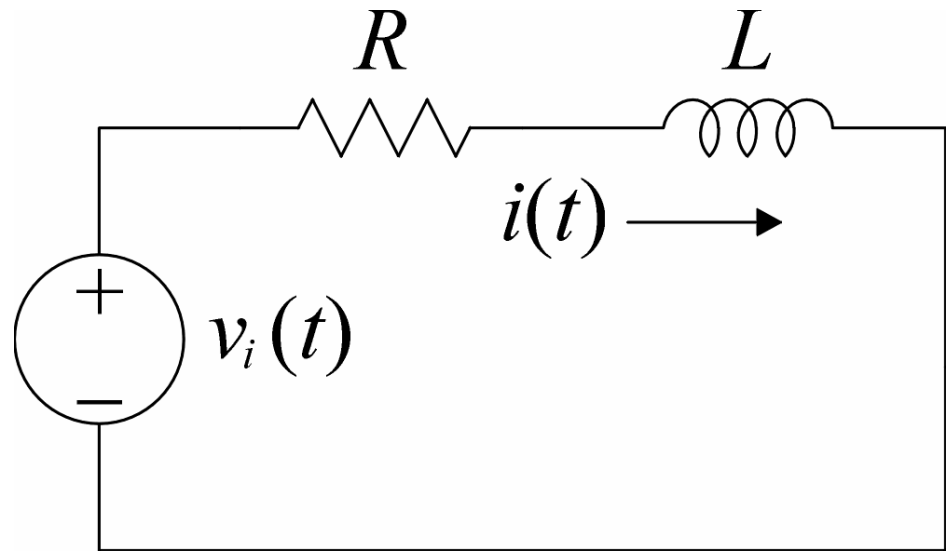
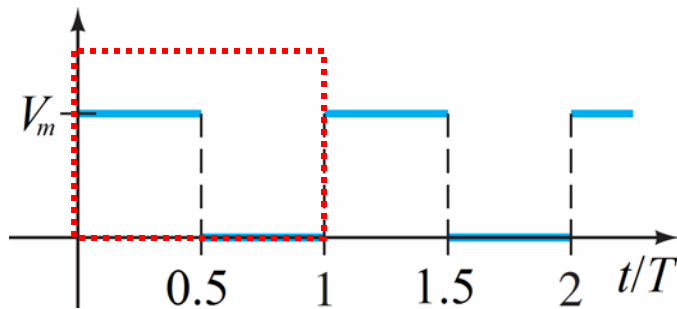
$$\Rightarrow (1 - e^{-sT}) I_2 = e^{-sT} I_1, \quad I_2 = \frac{e^{-sT}}{1 - e^{-sT}} I_1;$$

$$\Rightarrow F(s) = I_1 + \frac{e^{-sT}}{1 - e^{-sT}} I_1 = \frac{1}{1 - e^{-sT}} I_1.$$

Example 8 (1)

- An RL circuit driven by a periodic square voltage $v_i(t)$:

$$\begin{cases} Li'(t) + Ri(t) = v_i(t); \\ i(0) = I_0 \end{cases}$$



$$V_i(s) = \frac{1}{1 - e^{-Ts}} \int_0^{T/2} V_m \times e^{-st} dt = \frac{V_m}{1 - e^{-Ts}} \times \frac{1 - e^{-Ts/2}}{s} = \frac{V_m}{s(1 + e^{-Ts/2})}.$$

Example 8 (2)

67

■ Step 1: $L\{Li'(t) + Ri(t)\} = L[sI(s) - I_0] + RI(s)$

$$= (Ls + R)I(s) - LI_0 = L\{v_i(t)\} = V_i(s) = \frac{V_m}{s(1 + e^{-Ts/2})}.$$

■ Step 2: $I(s) = \frac{I_0}{s + R/L} + \frac{V_m/L}{s(s + R/L)(1 + e^{-Ts/2})} \equiv I_1 + I_2.$

■ Step 3: $i_1(t) = L^{-1}\left\{\frac{I_0}{s + R/L}\right\} = I_0 e^{-t/\tau}$, where $\tau = \frac{L}{R}$.

$$I_2(s) = \frac{V_m}{L} \times \frac{1}{s(s + \tau^{-1})} \times \frac{1}{1 + e^{-Ts/2}} \equiv \frac{V_m}{L} \times G(s) \times H(s),$$

Example 8 (3)

68

$$G(s) = \frac{1}{s(s + \tau^{-1})} = \frac{\tau}{s} - \frac{\tau}{s + \tau^{-1}},$$

$$H(s) = \frac{1}{1 + e^{-Ts/2}} = 1 - e^{-\frac{T}{2}s} + e^{-Ts} - \dots,$$

$$I_2(s) = \frac{V_m}{L} \times \tau \left(\frac{1}{s} - \frac{1}{s + \tau^{-1}} \right) \times (1 - e^{-\frac{T}{2}s} + e^{-Ts} - \dots)$$

$$= I_m \left[\left(\frac{1}{s} - \frac{e^{-\frac{T}{2}s}}{s} + \frac{e^{-Ts}}{s} - \dots \right) - \left(\frac{1}{s + \tau^{-1}} - \frac{e^{-\frac{T}{2}s}}{s + \tau^{-1}} + \frac{e^{-Ts}}{s + \tau^{-1}} - \dots \right) \right], \quad I_m = \frac{V_m}{R}.$$

$$i_2(t) = I_m \left\{ [u(t) - u(t - T/2) + u(t - T) - \dots] \right. \\ \left. - [e^{-t/\tau} u(t) - e^{-(t-1)/\tau} u(t - T/2) + e^{-(t-2)/\tau} u(t - T) - \dots] \right\}$$

Example 8 (4)

$$i_2(t) = I_m \left\{ (1 - e^{-t/\tau}) \times u(t) - \left[1 - e^{-(t-T/2)/\tau} \right] \times u(t - T/2) + \left[1 - e^{-(t-T)/\tau} \right] \times u(t - T) - \dots \right\}$$

$$\Rightarrow i_2(t) = I_m \sum_{q=0}^n (-1)^q \left[1 - e^{-(t-qT/2)/\tau} \right], \text{ if } \frac{n}{2}T < t < \frac{n+1}{2}T.$$

■ For $t < T/2$ ($n = 0$): $i_2(t) = I_m (1 - e^{-t/\tau})$;

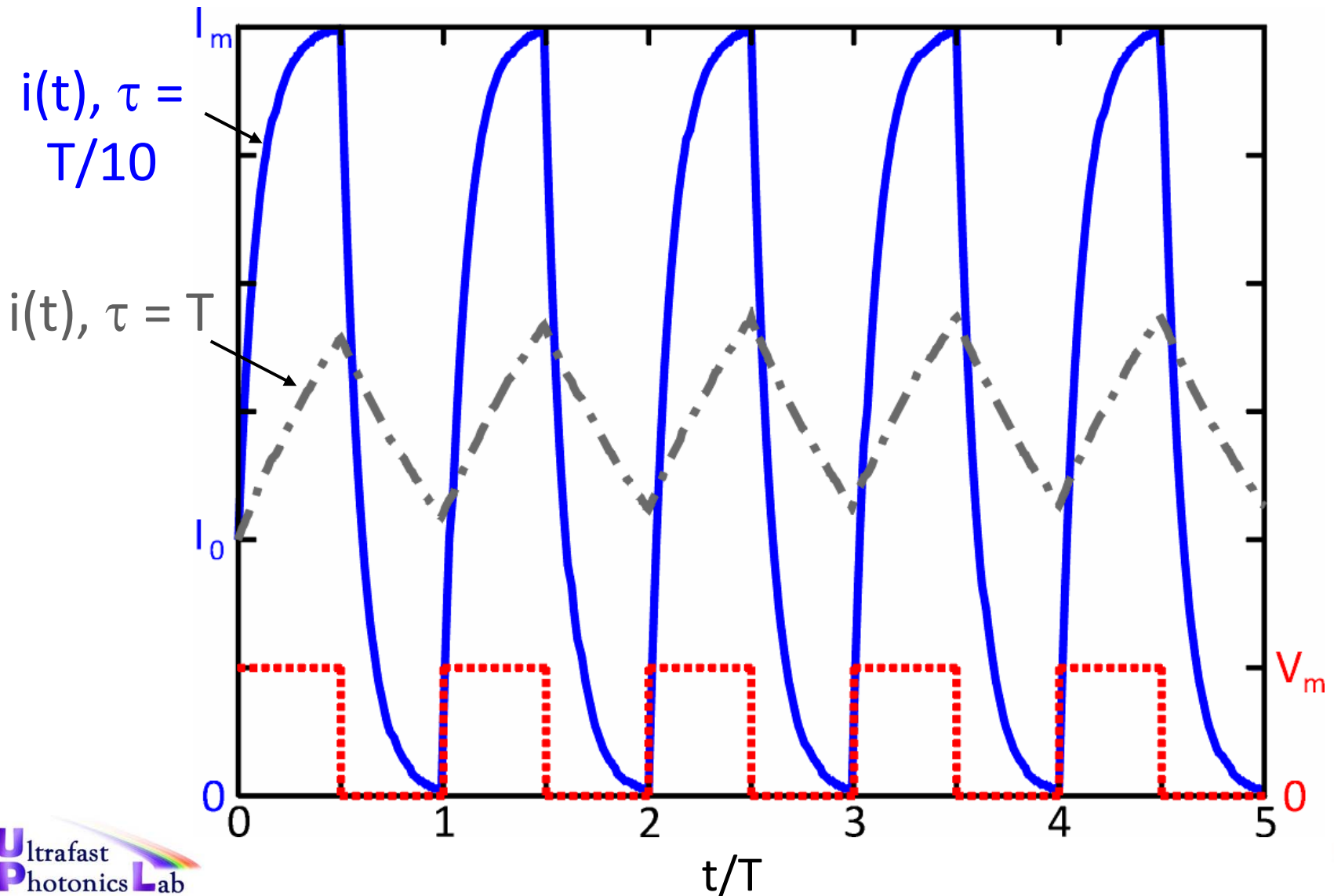
$$\Rightarrow i(t) = i_1(t) + i_2(t) = I_0 e^{-t/\tau} + I_m (1 - e^{-t/\tau}).$$

■ For $T/2 < t < T$ ($n = 1$): $i_2(t) = I_m \left\{ \cancel{(1 - e^{-t/\tau})} - \cancel{[1 - e^{-(t-T/2)/\tau}]} \right\}$
 $= I_m [-e^{-t/\tau} + e^{-(t-T/2)/\tau}];$

$$\Rightarrow i(t) = I_0 e^{-t/\tau} + I_m [-e^{-t/\tau} + e^{-(t-T/2)/\tau}].$$

Solution plot

70



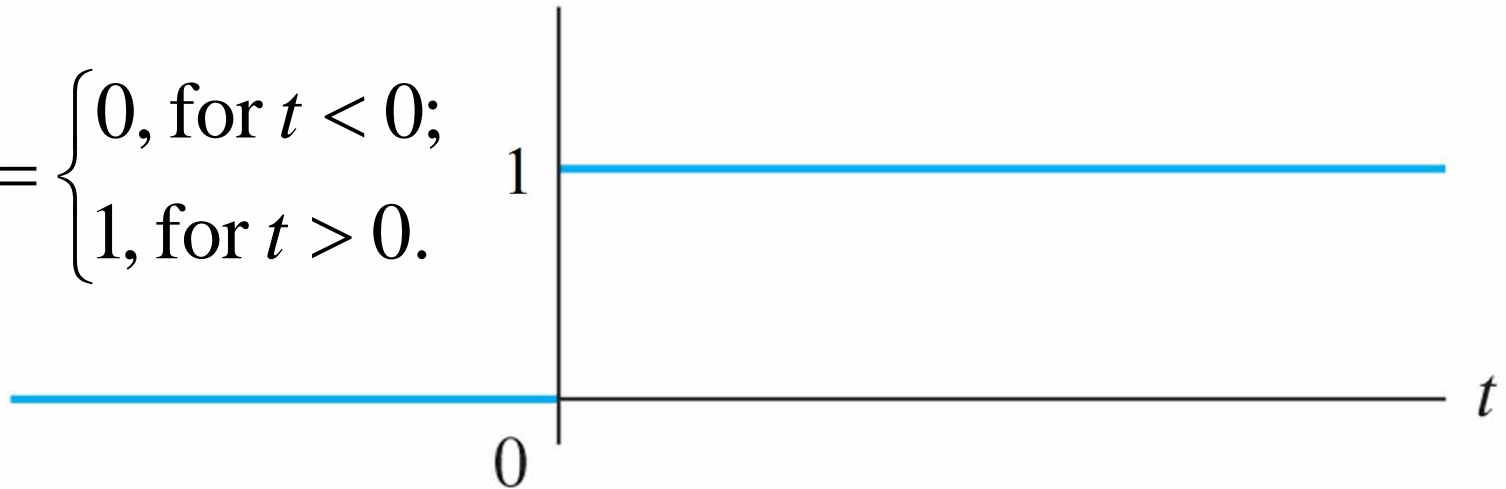
□ Dirac delta function $\delta(t)$

- Definition of $\delta(t)$
- $\delta'(t)$
- Impulse response $h(t)$

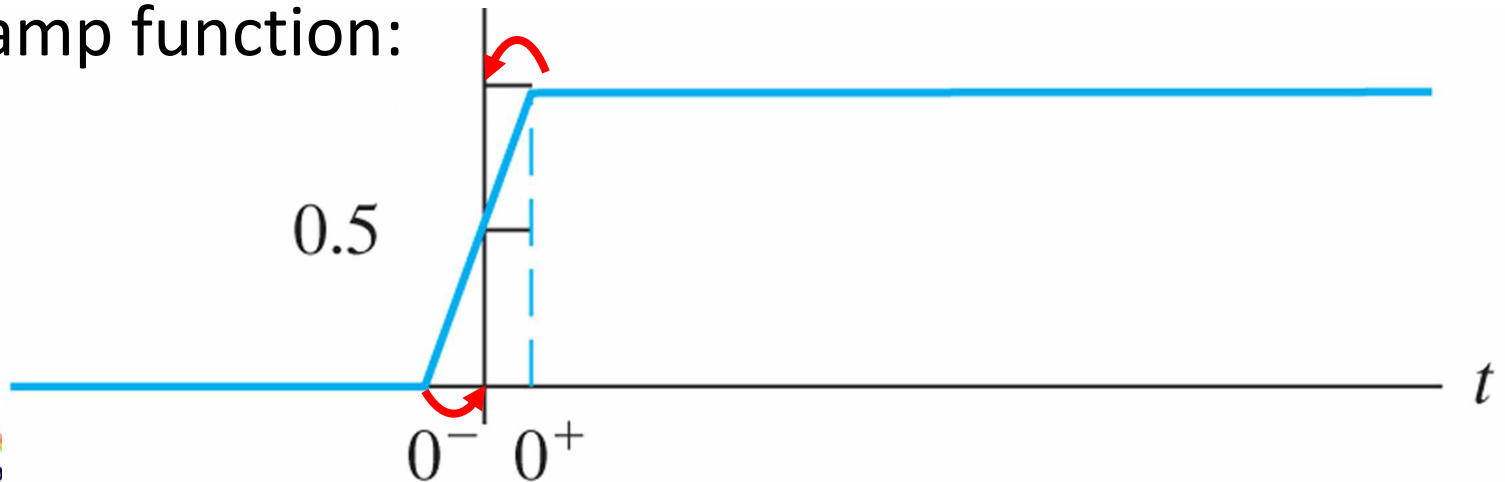
Revisit the unit-step function $u(t)$

72

$$u(t) = \begin{cases} 0, & \text{for } t < 0; \\ 1, & \text{for } t > 0. \end{cases}$$

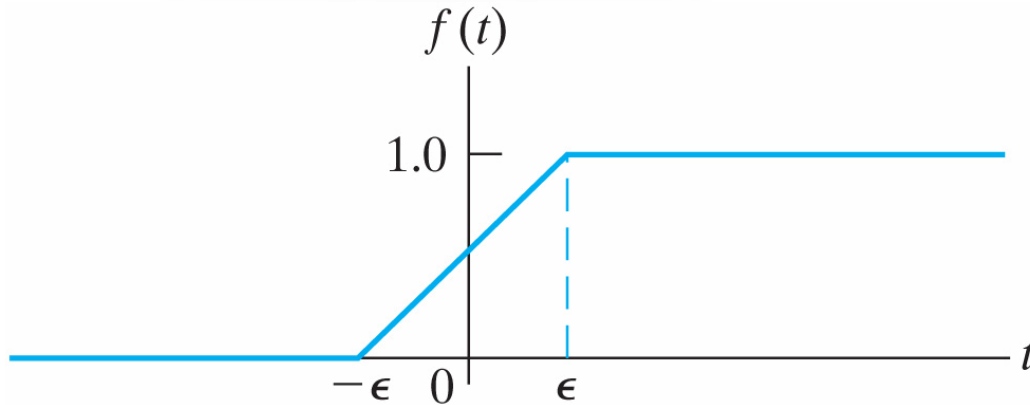


- $u(t)$ can be approximated by the limit of a linear ramp function:

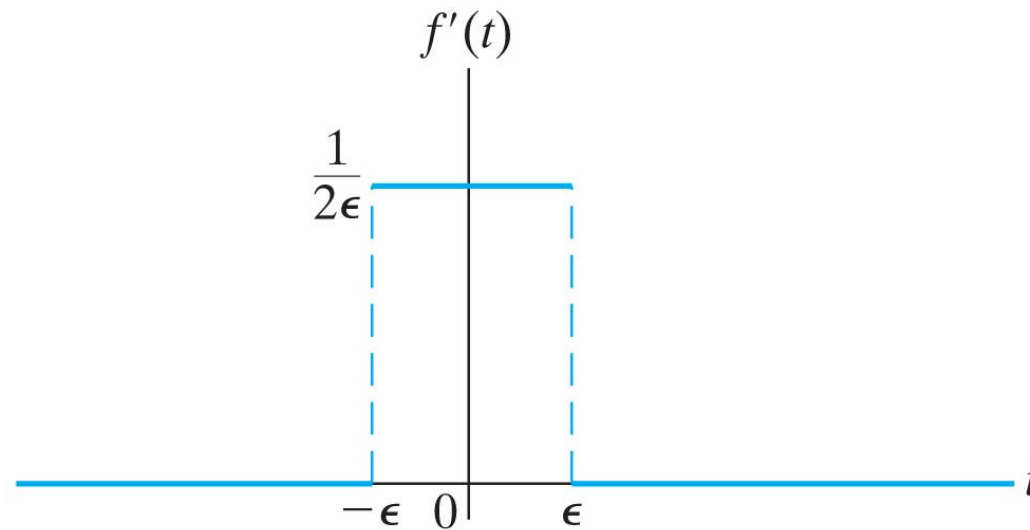


Impulse function $\delta(t)$

73



$$u(t) = \lim_{\epsilon \rightarrow 0} f(t)$$



$$\delta(t) = \lim_{\epsilon \rightarrow 0} f'(t) = u'(t)$$

$$\Rightarrow \begin{cases} \delta(t) = \begin{cases} \infty, & t = 0; \\ 0, & \text{otherwise;} \end{cases} \\ \int_{-\infty}^{\infty} \delta(t) dt = 1. \end{cases}$$

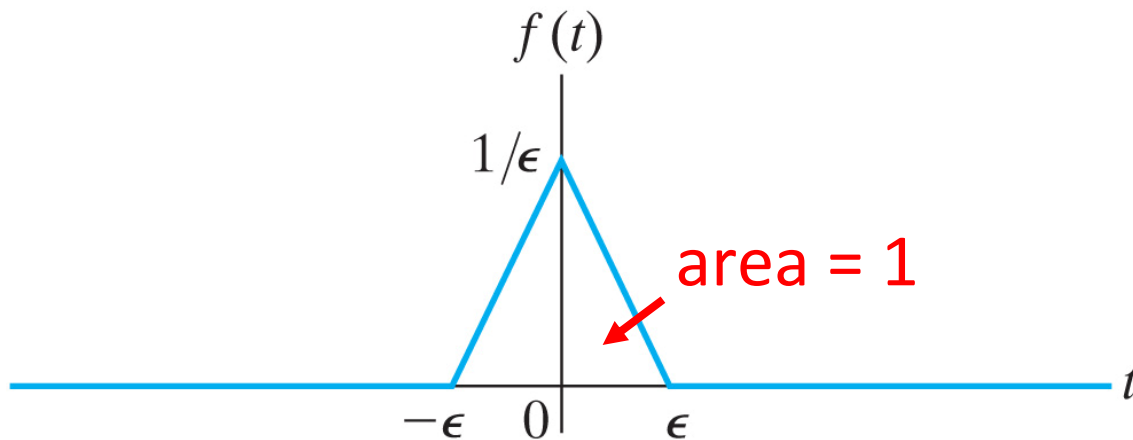
- **Sampling** of $f(t)$ at $t = a$ can be formulated by integral of the product of $f(t)$ and $\delta(t-a)$:

$$\begin{aligned}\int_0^\infty f(t) \times \delta(t-a) dt &= \lim_{\varepsilon \rightarrow 0} \int_{a-\varepsilon}^{a+\varepsilon} f(t) \times \delta(t-a) dt \\ &= f(a) \left[\lim_{\varepsilon \rightarrow 0} \int_{a-\varepsilon}^{a+\varepsilon} \delta(t-a) dt \right] = f(a) \times 1 = f(a).\end{aligned}$$

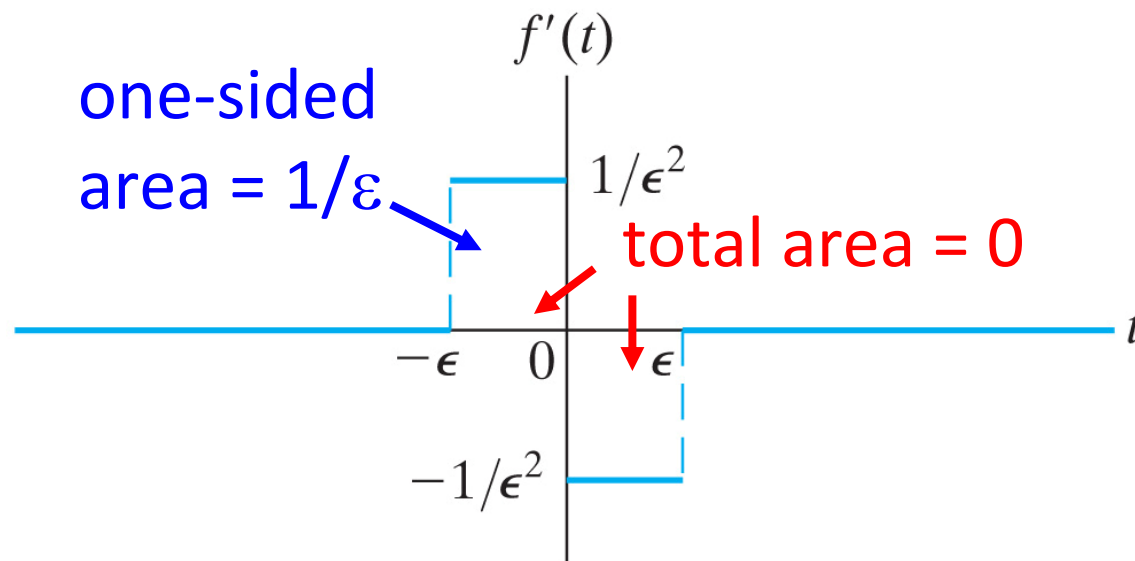
- It can be used in deriving **$L\{\delta(t)\} = 1$** :

$$L\{\delta(t)\} = \int_0^\infty \delta(t) \times e^{-st} dt = e^{-s(0)} = 1, \text{ for any } s \in C.$$

$\delta'(t) = ?$

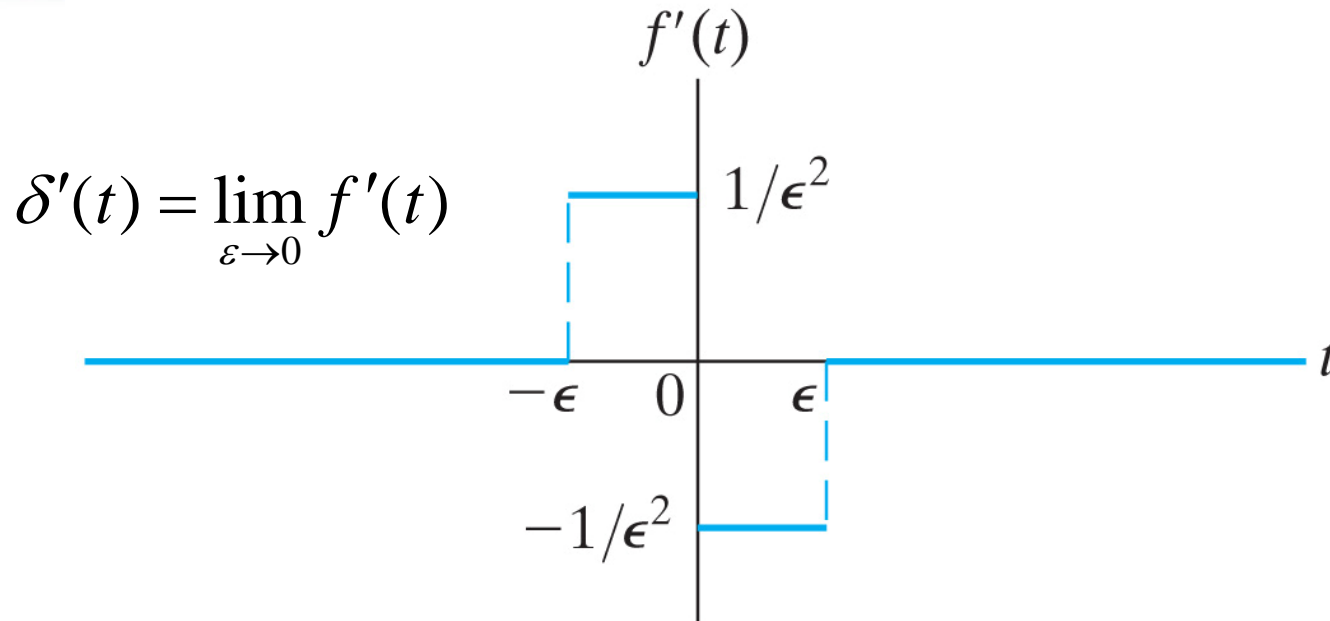


$$\delta(t) = \lim_{\epsilon \rightarrow 0} f(t)$$



$$\delta'(t) = \lim_{\epsilon \rightarrow 0} f'(t)$$

$L\{\delta'(t)\}$



$$\begin{aligned}
 L\{\delta'(t)\} &= \lim_{\epsilon \rightarrow 0} \int_{-\infty}^{\infty} f'(t) e^{-st} dt = \lim_{\epsilon \rightarrow 0} \left[\int_{-\epsilon}^{0^-} \frac{1}{\epsilon^2} e^{-st} dt + \int_{0^+}^{\epsilon} \left(\frac{-1}{\epsilon^2} \right) e^{-st} dt \right] \\
 &= \lim_{\epsilon \rightarrow 0} \frac{e^{s\epsilon} + e^{-s\epsilon} - 2}{s\epsilon^2} = \lim_{\epsilon \rightarrow 0} \frac{(e^{s\epsilon} - e^{-s\epsilon})s}{2s\epsilon} = \lim_{\epsilon \rightarrow 0} \frac{(e^{s\epsilon} + e^{-s\epsilon})s^2}{2s} = s.
 \end{aligned}$$

$L^{-1}\{\text{improper rational function}\}$

77

$$\begin{aligned} F(s) &= \frac{s^4 + 13s^3 + 66s^2 + 200s + 300}{s^2 + 4s + 20} \\ &= (s^2 + 4s + 10) - \frac{20}{s + 4} + \frac{50}{s + 5}. \end{aligned}$$

Red arrows indicate the mapping from the polynomial part and the partial fractions to the time-domain expression:

$$\begin{aligned} f(t) &= \delta''(t) + 4\delta'(t) + 10\delta(t) \\ &\quad + \left[-20e^{-4t} + 50e^{-5t} \right] u(t). \end{aligned}$$

Example 1 (1)

78

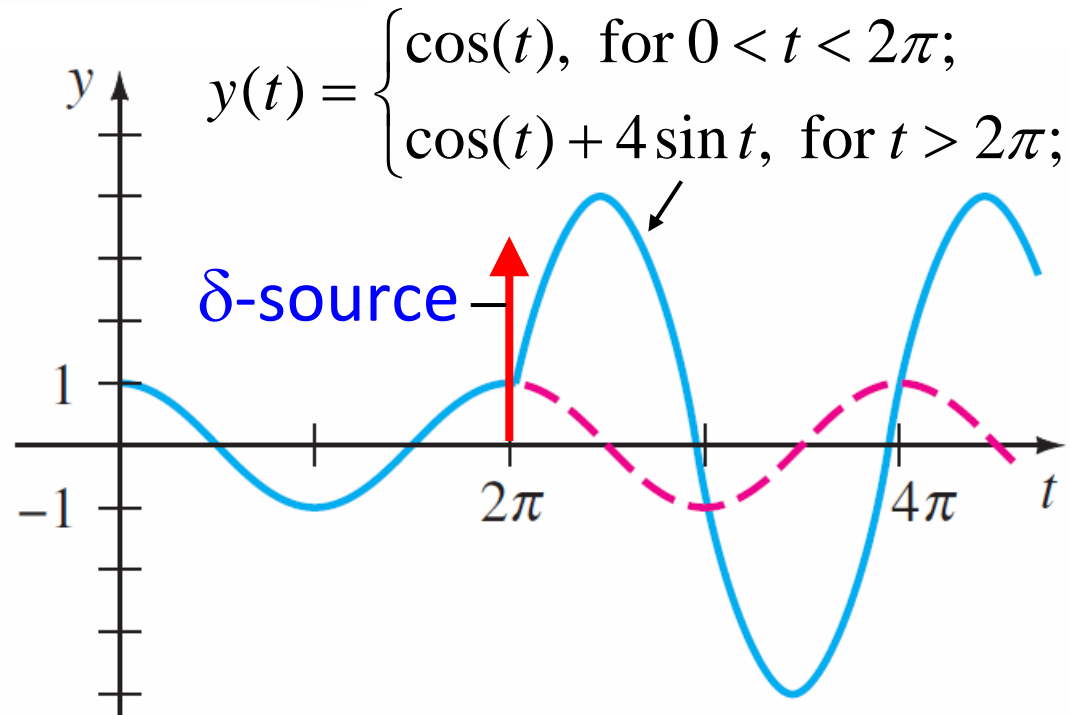
- ODE: $y'' + y = 4 \times \delta(t - 2\pi)$
- ICs: $y(0) = 1$, $y'(0) = 0$.
- Step 1: $L\{y'' + y\} = [s^2 Y(s) - s \times 1 - 0] + Y(s) = (s^2 + 1)Y(s) - s = L\{4 \times \delta(t - 2\pi)\} = 4 \times L\{\delta(t)\} \times e^{-2\pi s} = 4e^{-2\pi s}$.
- Step 2: $(s^2 + 1)Y(s) - s = 4e^{-2\pi s}, \Rightarrow$

$$Y(s) = \frac{s}{s^2 + 1^2} + \frac{4e^{-2\pi s}}{s^2 + 1^2} \equiv Y_1 + Y_2.$$

- Step 3: $y_1(t) = \cos(t)$; $y_2(t) = 4L^{-1}\{1/(s^2 + 1^2)\} \big|_{t \rightarrow t - 2\pi} = 4\sin(t - 2\pi) \times u(t - 2\pi) = 4\sin(t) \times u(t - 2\pi)$.

Example 1 (2)

79



- An impulse source to an LC circuit or mass-spring system could cause **discontinuous slope** but not necessarily discontinuous function value.

Impulse response

80

- ODE: $y'' + Py' + Qy = \delta(t)$
- ICs: $y(0) = 0$, $y'(0) = 0$.
- Step 1: $L\{y'' + Py' + Qy\} = [s^2Y(s) - s \times 0 - 0] + P[sY(s) - 0] + Q \times Y(s) = (s^2 + Ps + Q)Y(s) = L\{\delta(t)\} = 1$.
- Step 2: $(s^2 + Ps + Q)Y(s) \equiv D(s) \times Y(s) = 1, \Rightarrow Y(s) \equiv H(s) = 1/D(s)$.
- Step 3: $y(t) \equiv h(t) = L^{-1}\{1/D(s)\}$, which is named the “impulse response” (the output as a result of impulse source).

Why is $h(t)$ important ?

81

- For an arbitrary source $f(t)$, the output function $y(t)$ can be calculated by $h*f$:
- Step 1: $L\{y'' + Py' + Qy\} = D(s) \times Y(s) = L\{f(t)\} = F(s)$.
- Step 2: $Y(s) = F(s)/D(s) = H(s) \times F(s)$.
- Step 3: $y(t) = L^{-1}\{H(s) \times F(s)\} = h(t)*f(t)$, which is the **convolution** between the impulse response and the source function.
- A linear system is fully characterized by $h(t)$ or $H(s)$.

Summary

82

- How to calculate $L\{f(t)\}$ and $L^{-1}\{F(s)\}$?
- How to solve ODEs by Laplace transform?
- What're the complementary operations between time- and s-domains?
- What really happens in a resonantly driven system?
- What's $\delta(t)$? What's $F\{\delta(t)\}$?
- What's impulse response $h(t)$? Why is it useful?