



# Lesson 3

## Transient Response of Transmission Lines

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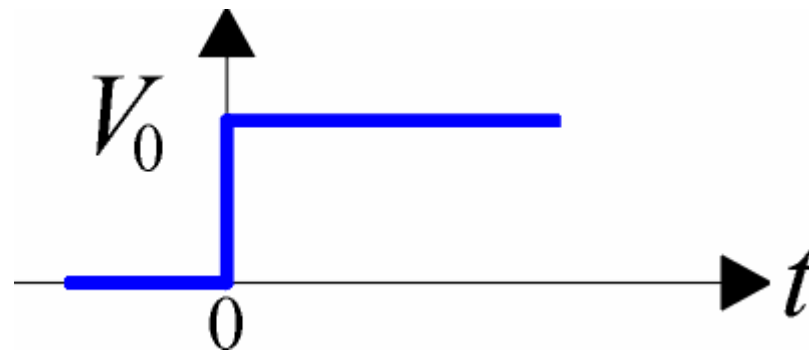
Institute of Photonics Technologies

Department of Electrical Engineering

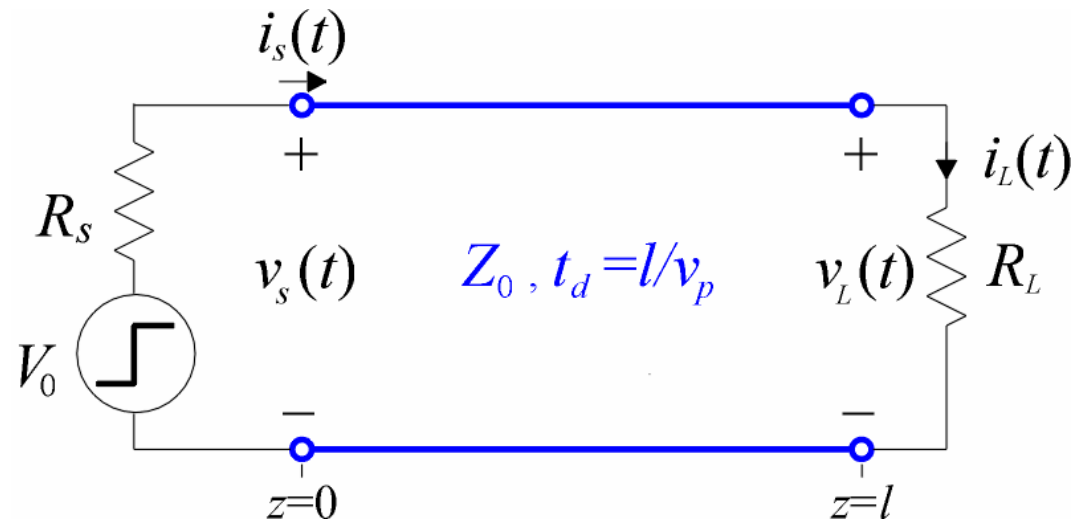
National Tsing Hua University, Taiwan

## Introduction

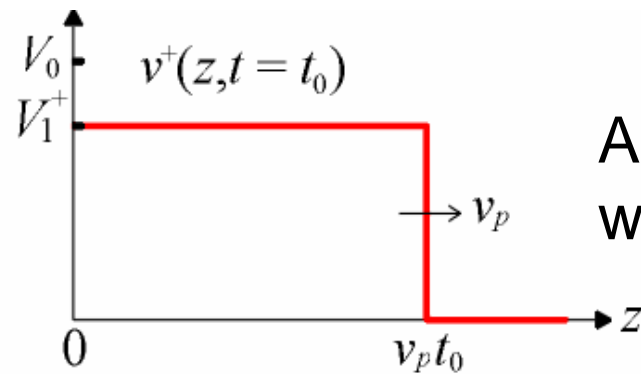
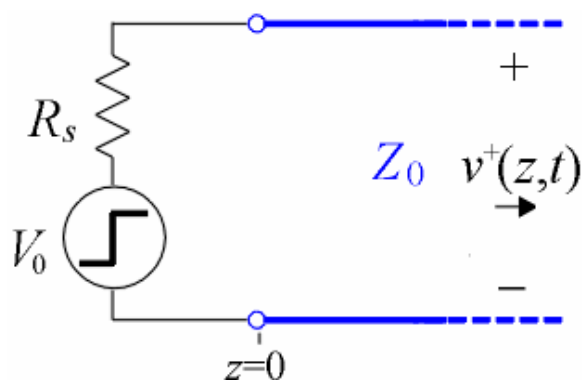
- When there is an interface between different materials ( $\varepsilon$ ,  $\mu$ ), discontinuity exists,  $\Rightarrow$  partial reflection & transmission,  $\Rightarrow$  total  $v$ ,  $i$  are determined by superposition (why?).
- Goal: Transient response of a terminated transmission line or cascaded lines excited by a **step-like** voltage source.



### Example 3-1: A finite line with resistive load-1



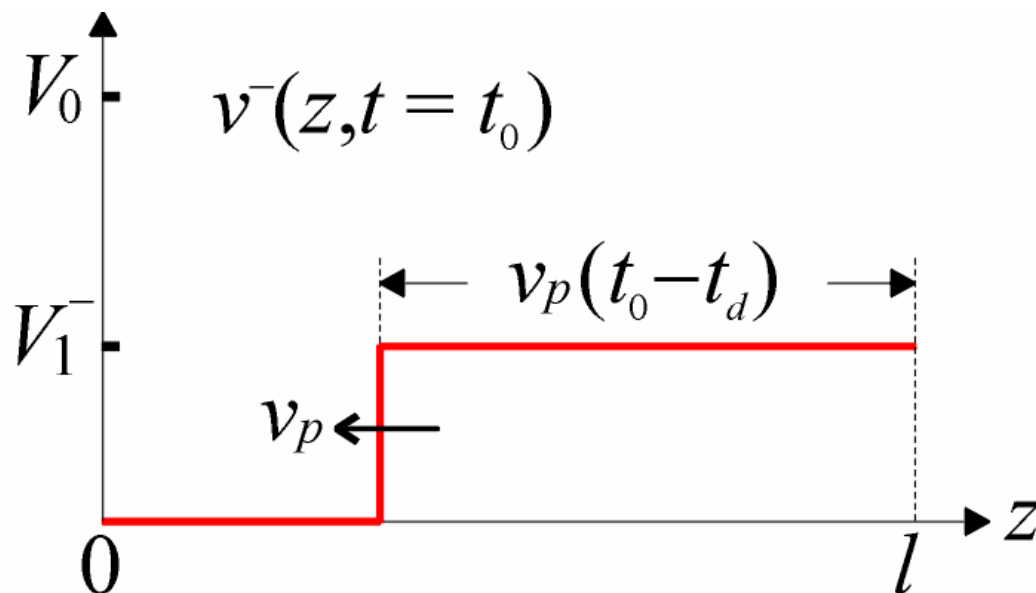
$$(1) \quad t > 0: v_1^+(z, t) = \begin{cases} V_1^+, & \text{if } z < v_p t \\ 0, & \text{otherwise} \end{cases}; \quad V_1^+ = \frac{Z_0}{Z_0 + R_s} V_0$$



As if there were no load.

### Example 3-1: A finite line with resistive load-2

$$(2) \ t > t_d: \ v_1^-(z, t) = \begin{cases} V_1^-, & \text{if } z > l - v_p(t - t_d) \\ 0, & \text{otherwise} \end{cases}$$



Reflected wave  $v_1^-(z, t)$  must arise, otherwise,

$$\frac{v_L(t)}{i_L(t)} = Z_0 \neq R_L,$$

in violation with the BC.

Meanwhile,  $v_1^+(z, t)$  acts as if it had kept on going to the right.



### Example 3-1: A finite line with resistive load-3

$$\left\{ \begin{array}{l} v_L(t) = v_1^+(l, t) + v_1^-(l, t) \\ i_L(t) = \frac{v_1^+(l, t)}{Z_0} - \frac{v_1^-(l, t)}{Z_0} \\ \text{BC: } v_L(t) = i_L(t) \cdot R_L(t) \end{array} \right.$$

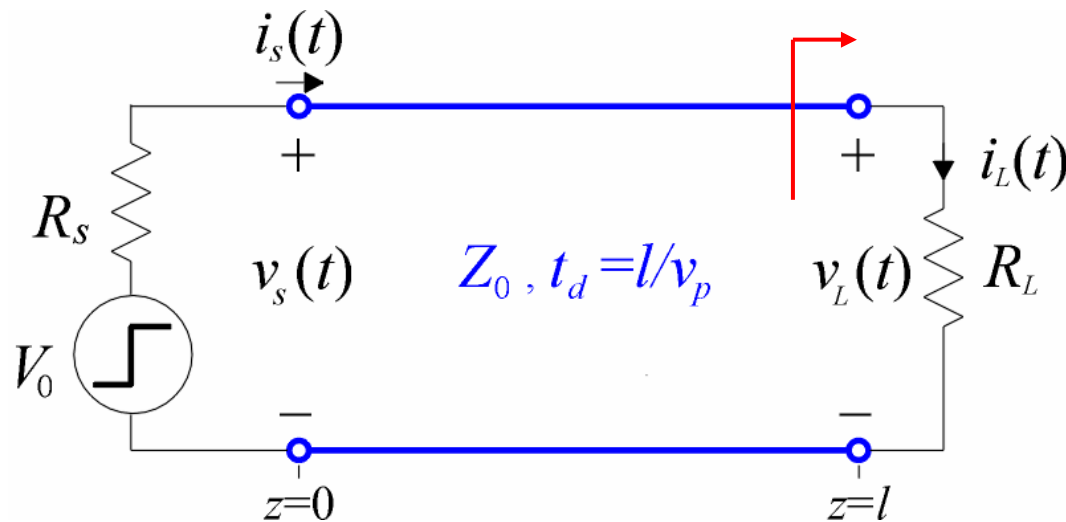
$$\Rightarrow v_1^+(l, t) + v_1^-(l, t) = \frac{R_L}{Z_0} [v_1^+(l, t) - v_1^-(l, t)]$$

$$\Rightarrow 1 + \frac{v_1^-(l, t)}{\underline{v_1^+(l, t)}} = \frac{R_L}{Z_0} \left[ 1 - \frac{v_1^-(l, t)}{\underline{v_1^+(l, t)}} \right]$$

### Example 3-1: A finite line with resistive load-4

$$\Rightarrow 1 + \Gamma_L = \frac{R_L}{Z_0} (1 - \Gamma_L)$$

$$\Gamma_L \equiv \frac{v_1^-(l, t)}{v_1^+(l, t)} = \frac{R_L - Z_0}{R_L + Z_0} \quad \text{Load voltage reflection coefficient}$$

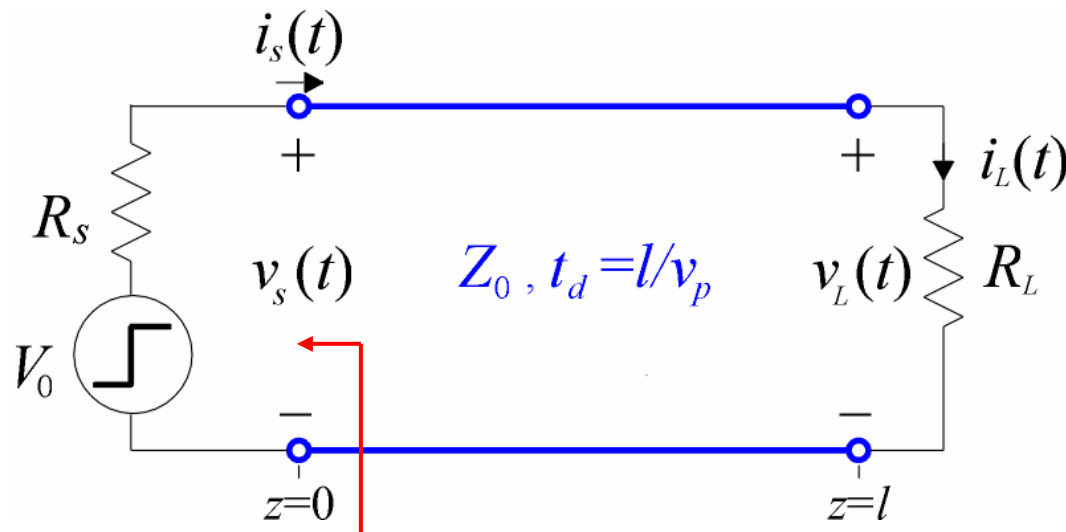


$\Gamma_L$  can be used to calculate  $V_1^-$  given  $V_1^+$  is known.

### Example 3-1: A finite line with resistive load-5

$$(3) \ t > 2t_d: \ v_2^+(z, t) = \begin{cases} V_2^+, & \text{if } z < v_p(t - 2t_d) \\ 0, & \text{otherwise} \end{cases}$$

$$\Gamma_s \equiv \frac{v_2^+(l, t)}{v_1^-(l, t)} = \frac{R_s - Z_0}{R_s + Z_0} \quad \text{Source voltage reflection coefficient}$$



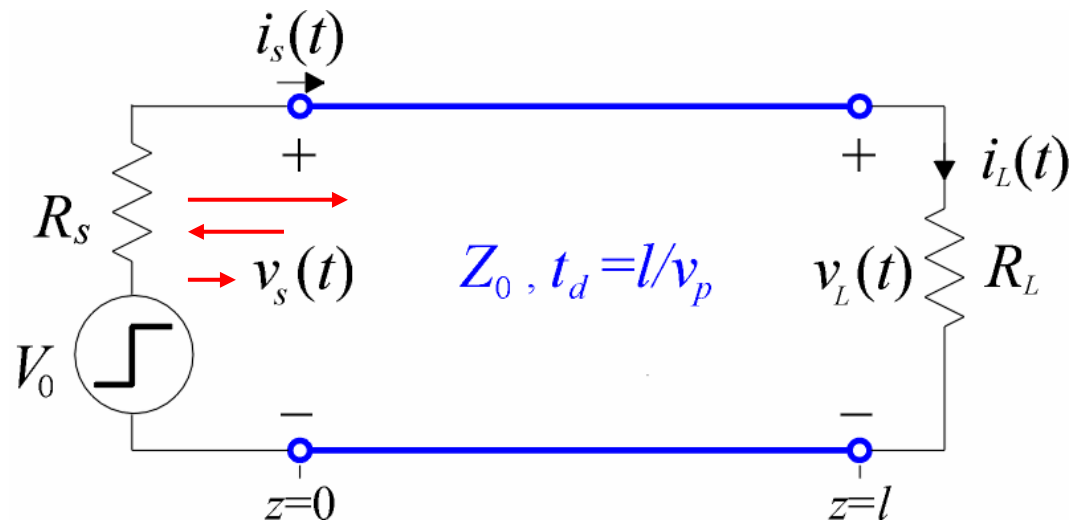
$\Gamma_s$  can be used to calculate  $V_2^+$  given  $V_1^-$  is known.

### Example 3-1: A finite line with resistive load-6

At  $t = 2t_d$  :

$$v_S(2t_d) = v_1^+(0, 2t_d) + v_1^-(0, 2t_d) + v_2^+(0, 2t_d) = v_1^+(0, 2t_d)(1 + \Gamma_L + \Gamma_L \Gamma_S)$$

$$i_S(2t_d) = \frac{v_1^+(0, 2t_d)}{Z_0} \ominus \frac{v_1^-(0, 2t_d)}{Z_0} \oplus \frac{v_2^+(0, 2t_d)}{Z_0} = \frac{v_1^+(0, 2t_d)}{Z_0} (1 - \Gamma_L + \Gamma_L \Gamma_S)$$





### Example 3-1: A finite line with resistive load-7

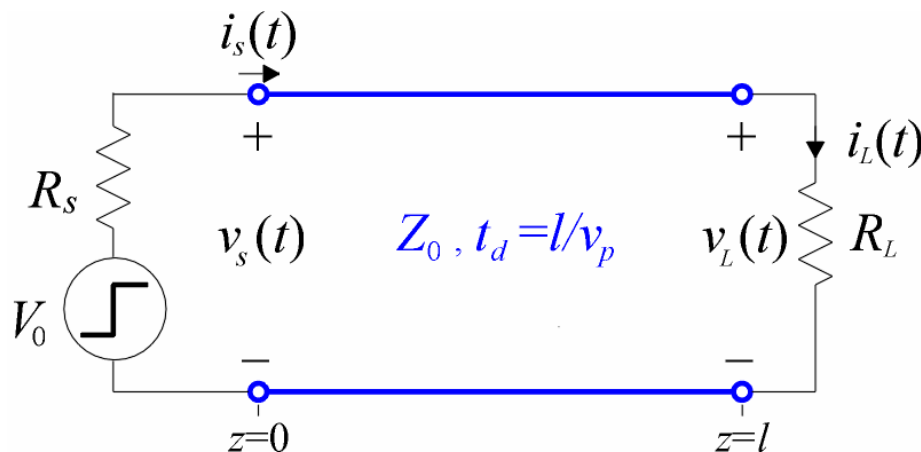
At  $t \rightarrow \infty$  :

$$\lim_{t \rightarrow \infty} v_s(t) = v_1^+(0, t \rightarrow \infty) \cdot \left( \underline{1 + \Gamma_L + \Gamma_L \Gamma_S + \Gamma_L^2 \Gamma_S^2 + \Gamma_L^2 \Gamma_S^2 + \Gamma_L^3 \Gamma_S^2 + \dots} \right)$$

$$= V_1^+ \cdot \left[ (1 + \Gamma_L) + \Gamma_L \Gamma_S (1 + \Gamma_L) + \Gamma_L^2 \Gamma_S^2 (1 + \Gamma_L) + \dots \right] = \underbrace{V_1^+}_{\downarrow} \frac{1 + \Gamma_L}{1 - \Gamma_L \Gamma_S}$$

$$\Rightarrow \lim_{t \rightarrow \infty} v_s(t) = \frac{R_L}{R_S + R_L} V_0$$

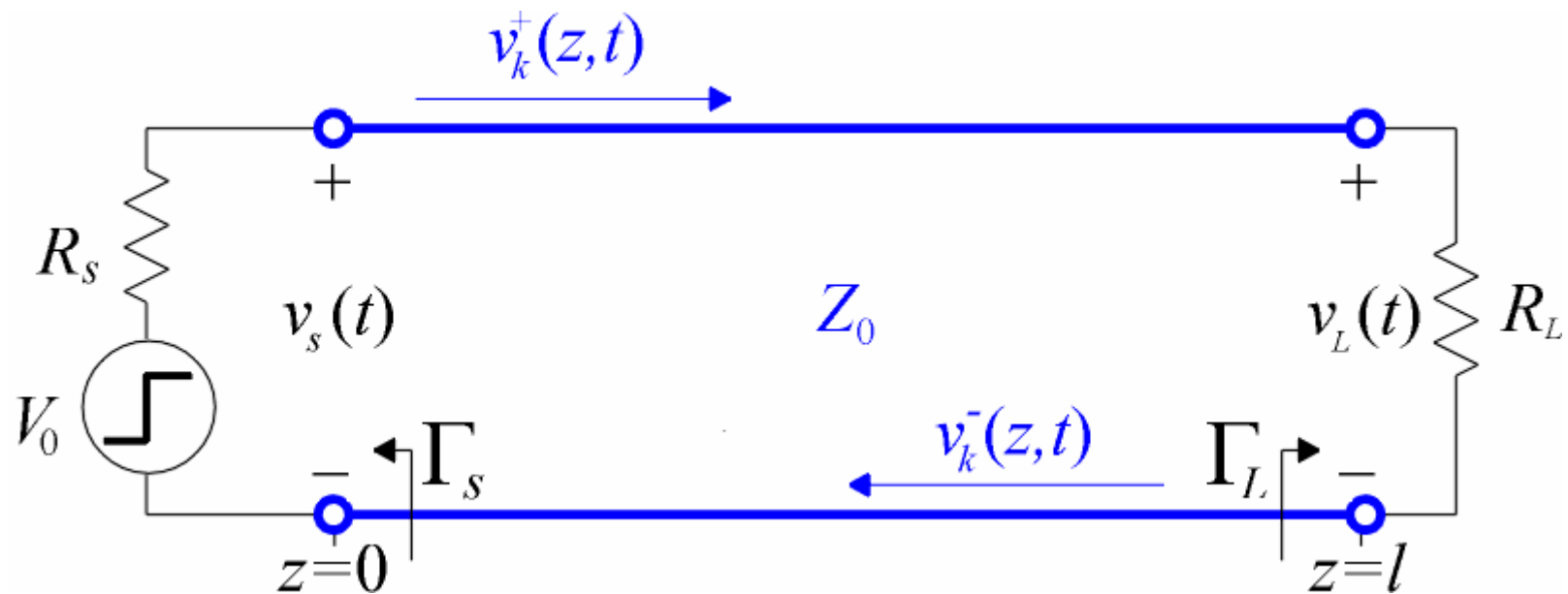
$$V_1^+ = \frac{Z_0}{Z_0 + R_S} V_0$$



Steady state response is as if there were no line.

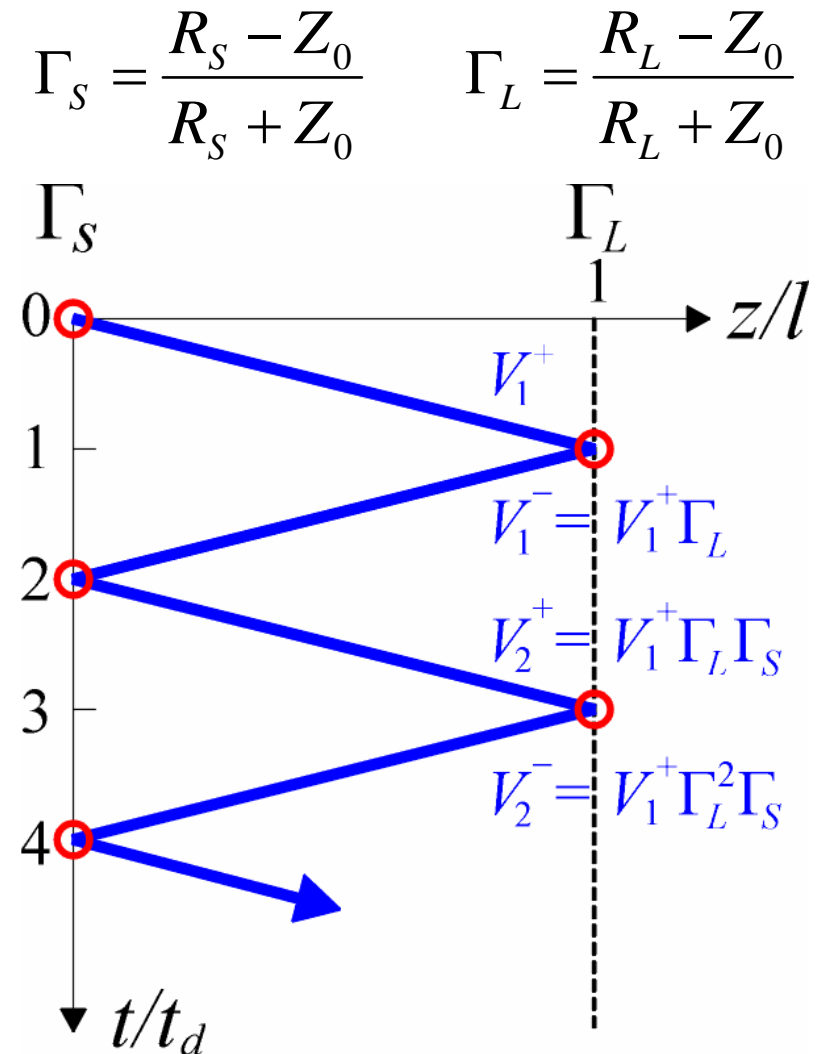
What is the usefulness of bounce diagram?

- Bounce diagram is a convenient tool to solve for transient response of a TX line driven by a **step voltage** and terminated by a **resistor**.



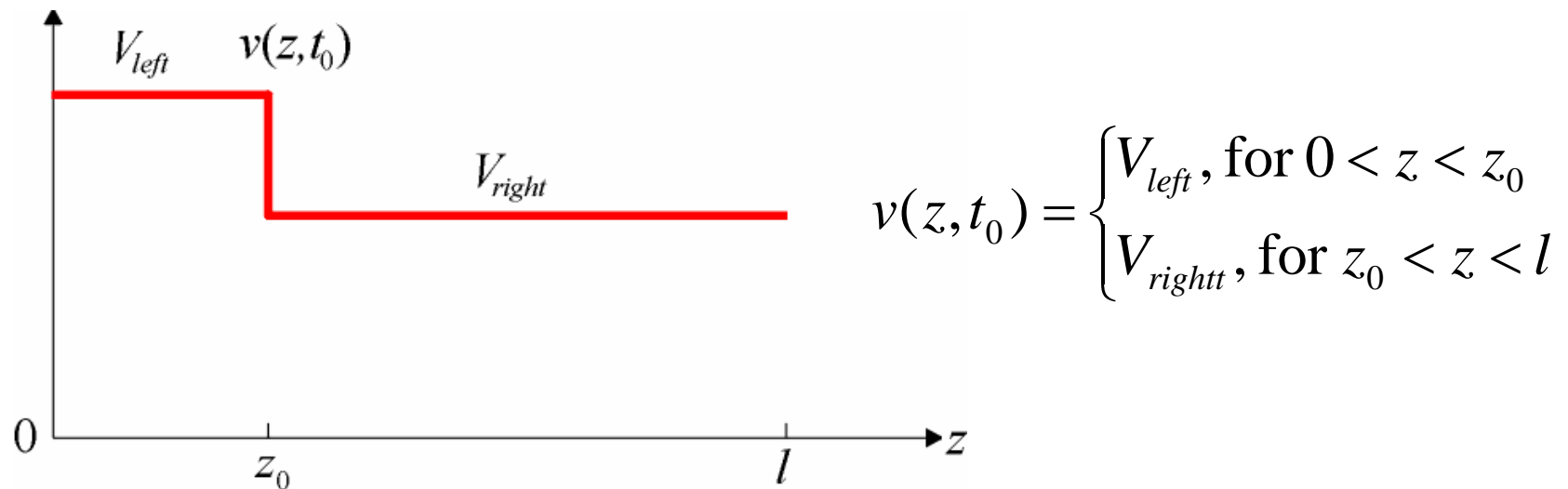
## How to draw a bounce diagram?

- Draw a 2D window with  $0 < z/l < 1$ ,  $t/t_d > 0$ . Denote the two reflection coefficients  $\Gamma_S$ ,  $\Gamma_L$ .
- Mark the points  $(0, 2n)$  and  $(1, 2n+1)$  for  $n = 0, 1, 2, \dots$
- Connecting the points by zigzag lines.
- Mark the voltage amplitude of each component wave.



Determine the spatial voltage distribution at some time instant-1

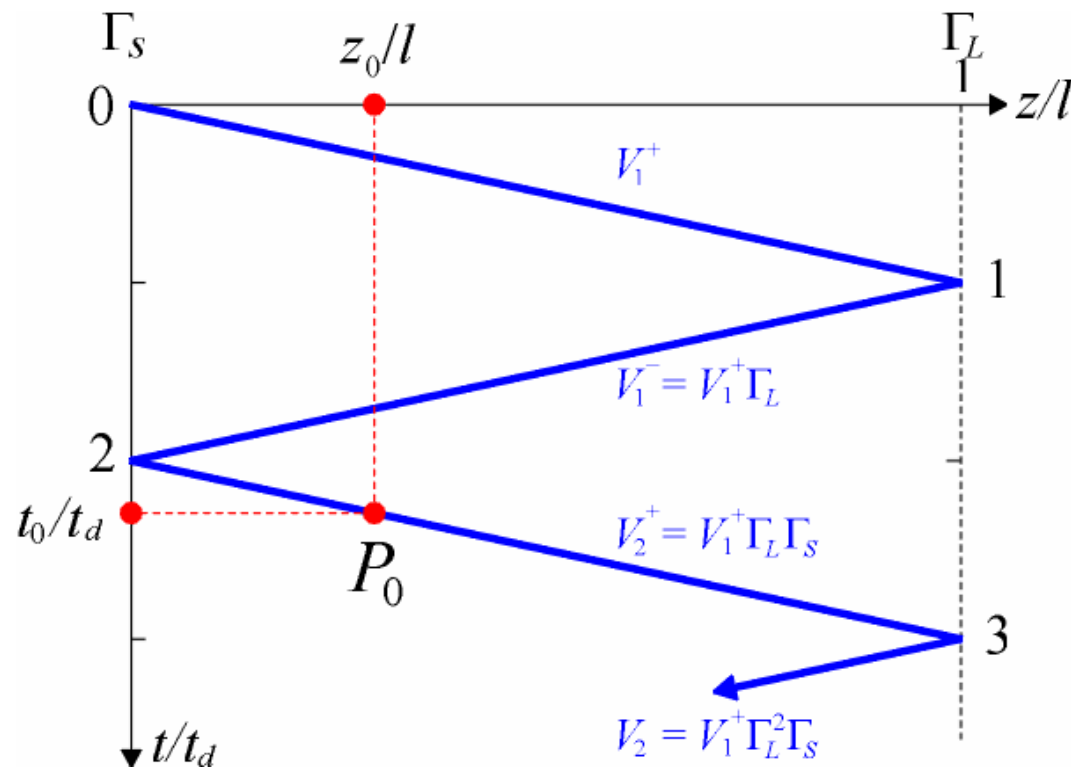
- The spatial distribution must be of the form:



The problem is how to find the three key parameters:  $z_0$ ,  $V_{left}$ , and  $V_{right}$ , respectively.

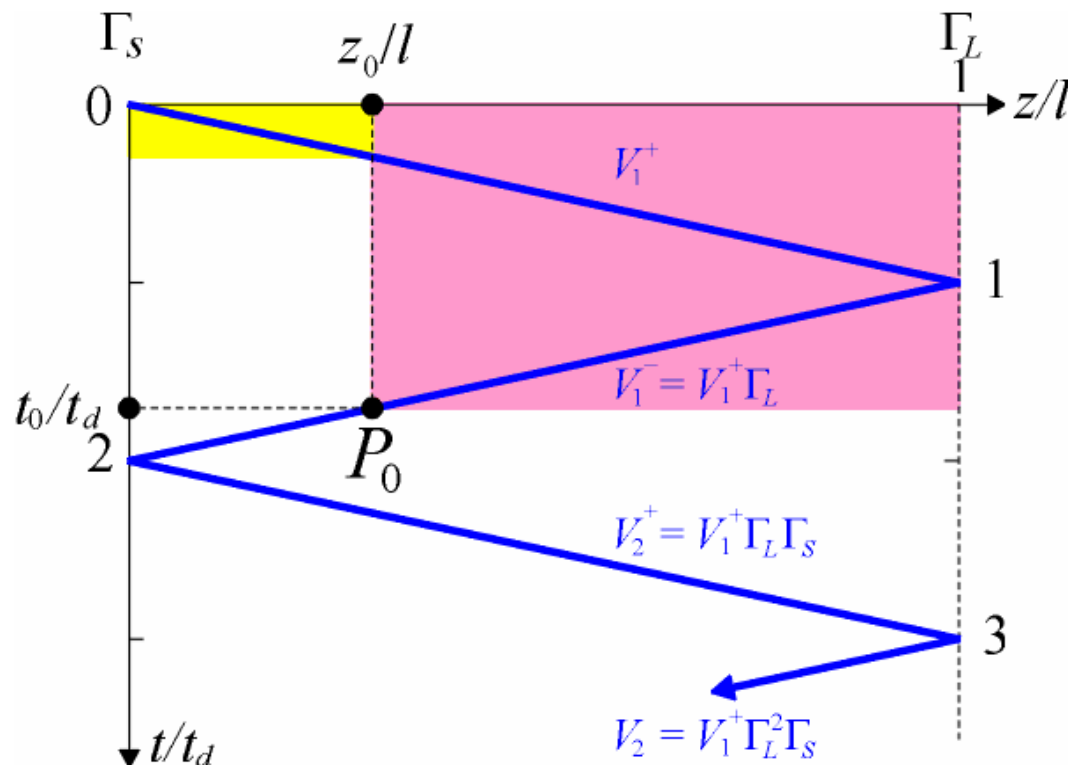
Determine the spatial voltage distribution at some time instant-2

- Draw a horizontal line  $t = t_0$  (i.e.  $t/t_d = t_0/t_d$ ), intersecting with some zigzag line at point  $P_0$  with coordinate  $(z_0, t_0)$ .



Determine the spatial voltage distribution at some time instant-3

- The determination of  $V_{left}$  and  $V_{right}$  has two cases.
- If  $P_0$  falls on a line with **positive** slope:



$$V_{left} = V_1^+$$

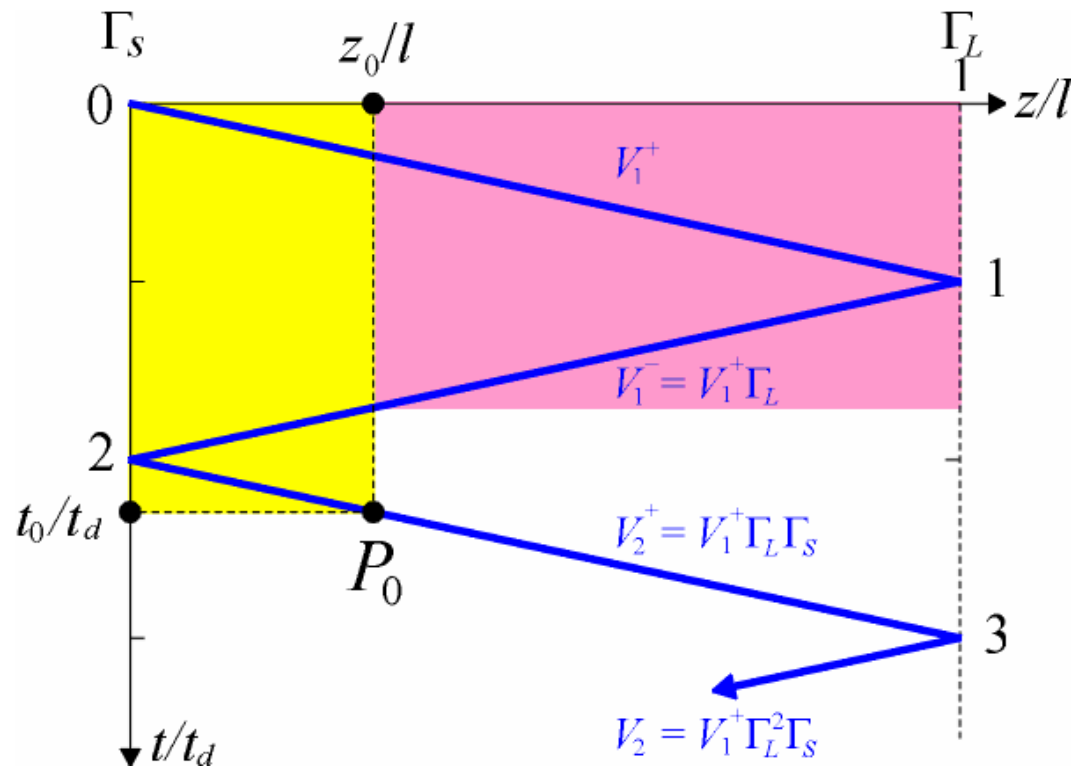
$$V_{right} = V_1^+ + V_1^-$$

$$= V_1^+ (1 + \Gamma_L)$$

$V_{right}$  has one more component.

Determine the spatial voltage distribution at some time instant-4

- If  $P_0$  falls on a line with **negative** slope:



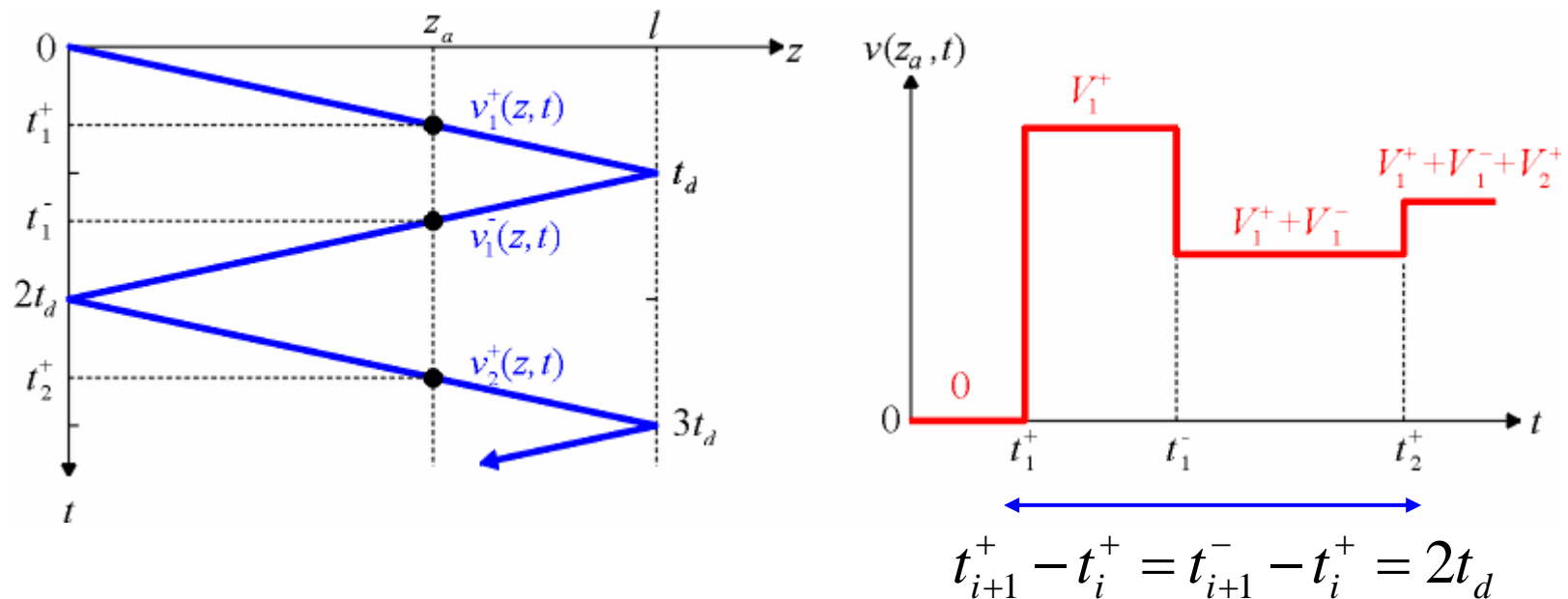
$$V_{left} = V_1^+ + V_1^- + V_2^+ \\ = V_1^+ (1 + \Gamma_L + \Gamma_L \Gamma_s)$$

$$V_{right} = V_1^+ + V_1^- \\ = V_1^+ (1 + \Gamma_L)$$

$V_{left}$  has one more component.

Determine the temporal voltage distribution at some position

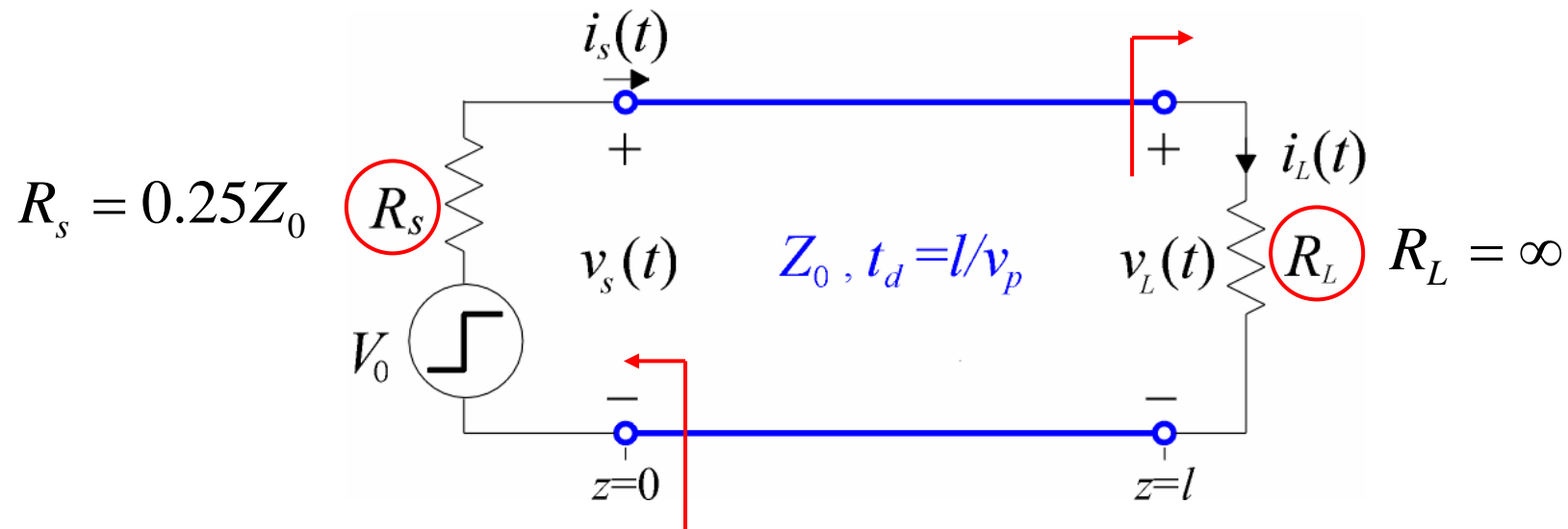
- $v(z_a, t)$  must have constant voltages switched at a series of time instants.
- Draw a vertical line  $z = z_a$ , intersecting with the zigzag lines at  $t = t_1^+, t_1^-, t_2^+, \dots$





### Example 3-2: Single TX line with open termination-1

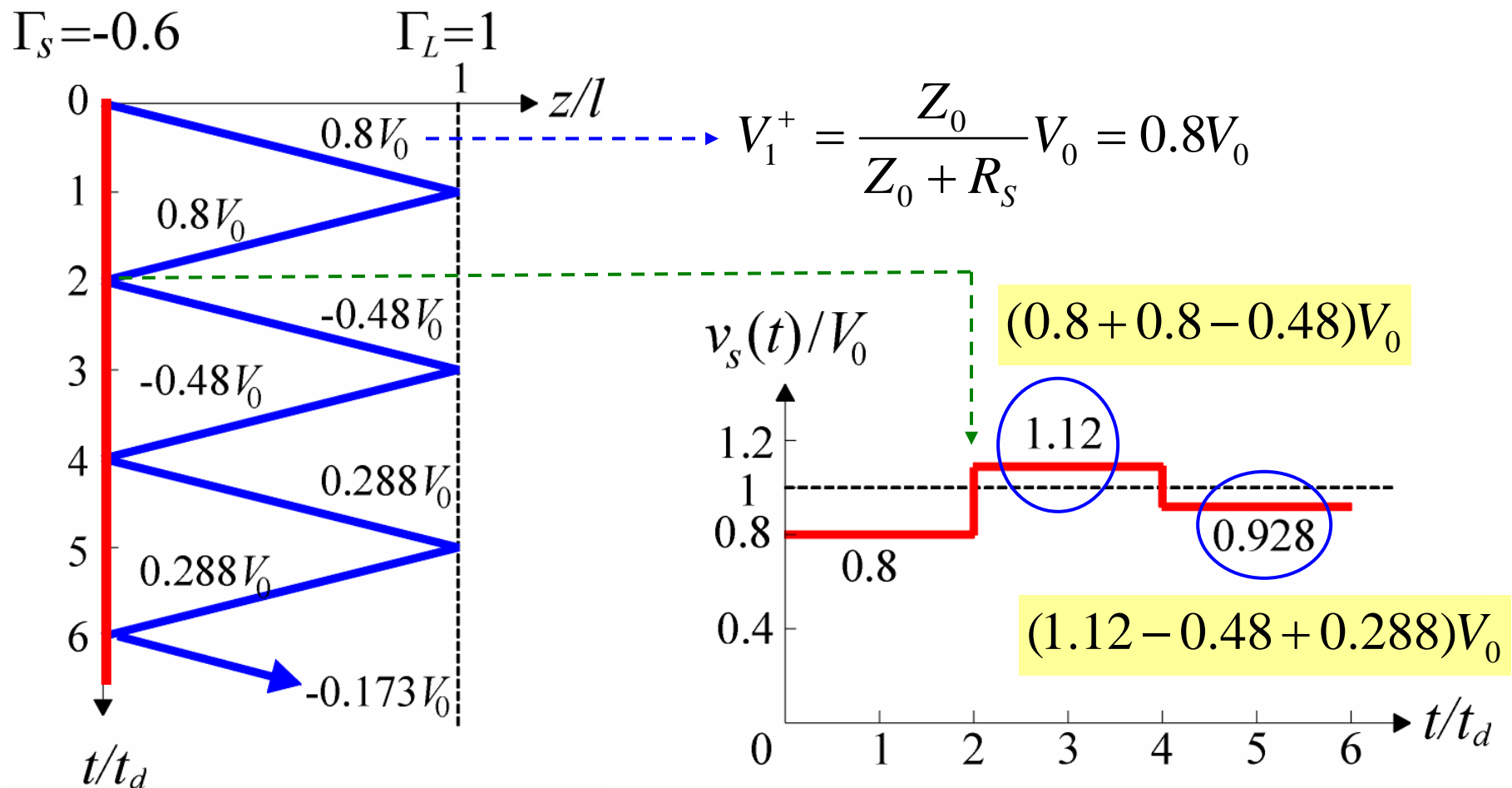
$$\Gamma_L = \frac{R_L - Z_0}{R_L + Z_0} = 1$$



$$\Gamma_s = \frac{R_s - Z_0}{R_s + Z_0} = -0.6$$

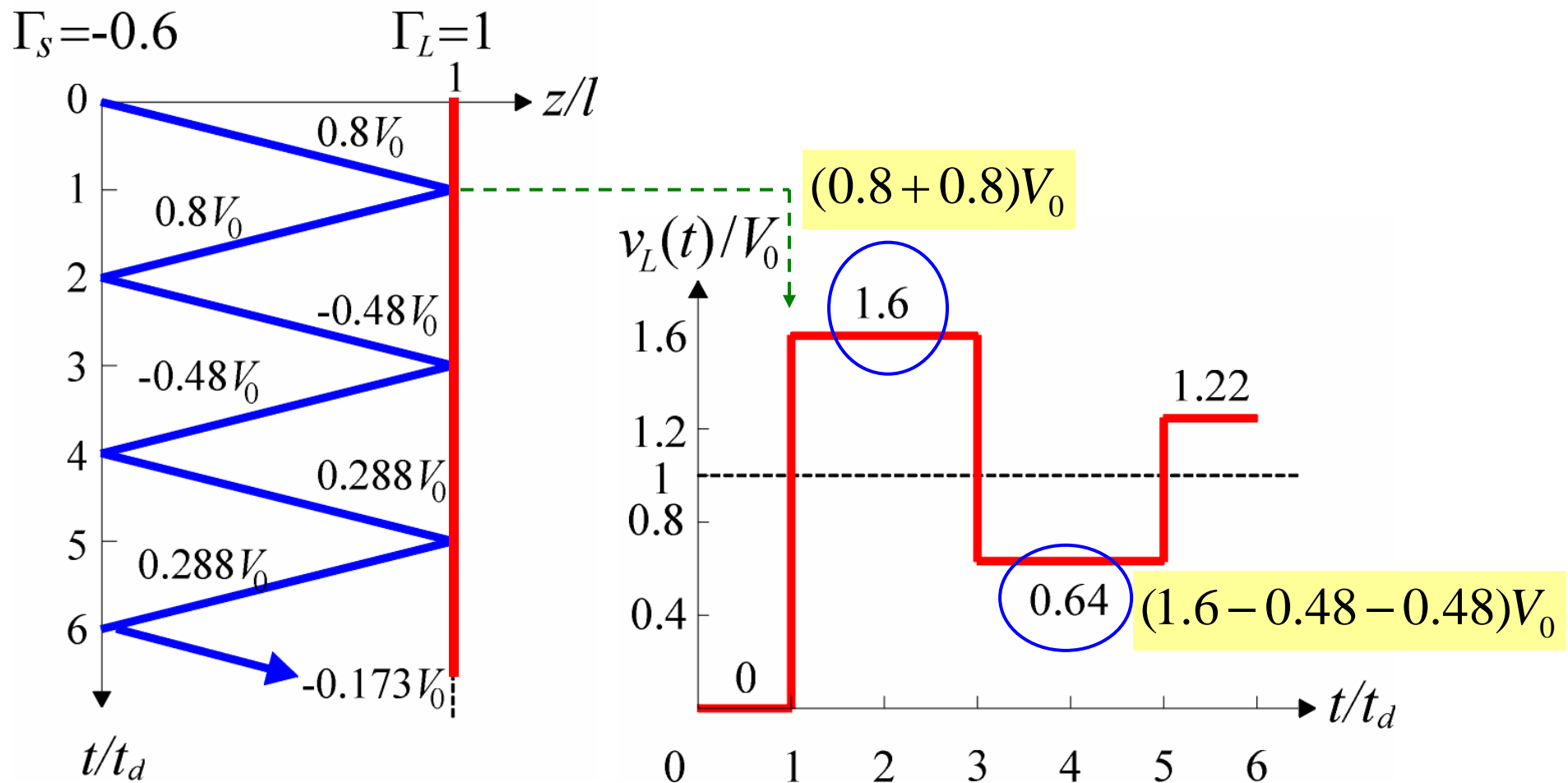
### Example 3-2: Single TX line with open termination-2

(1)  $z = 0$  :  $t_1^+ = 0, t_1^- = t_2^+ = 2t_d, t_2^- = t_3^+ = 4t_d, \dots$



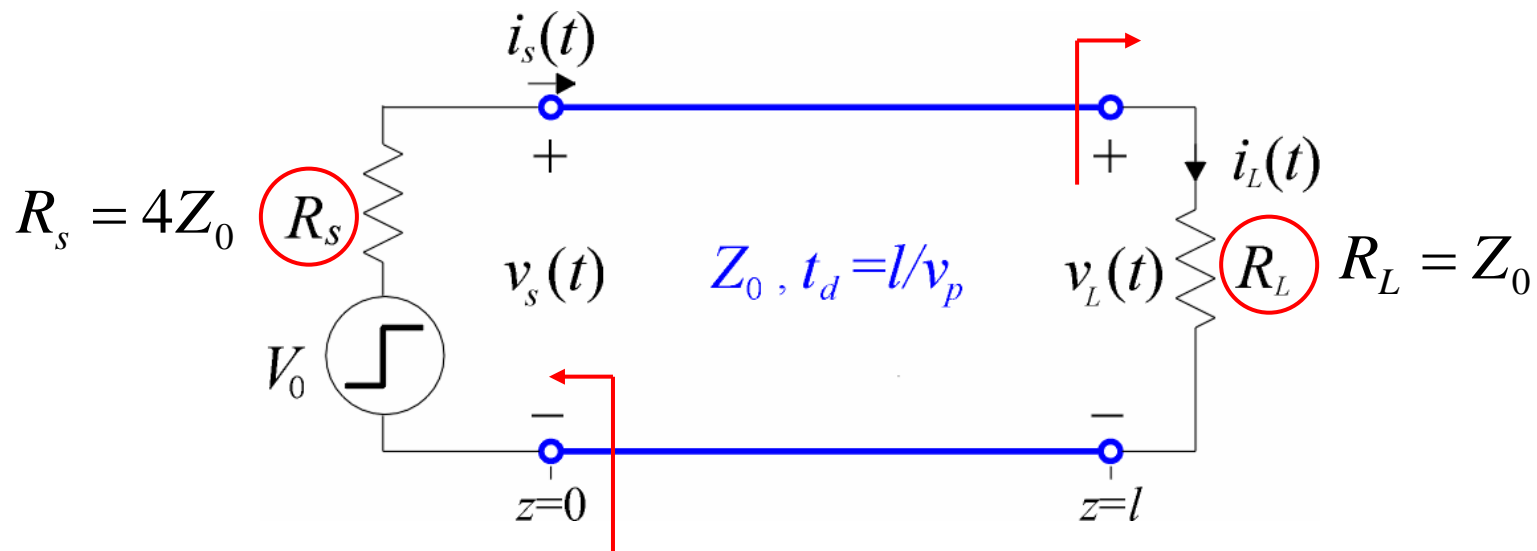
### Example 3-2: Single TX line with open termination-3

(2)  $z = l$  :  $t_1^+ = t_1^- = t_d$ ,  $t_k^+ = t_k^- = (2k - 1)t_d, \dots$



### Example 3-3: Single TX line with matched termination-1

$$\Gamma_L = \frac{R_L - Z_0}{R_L + Z_0} = 0$$

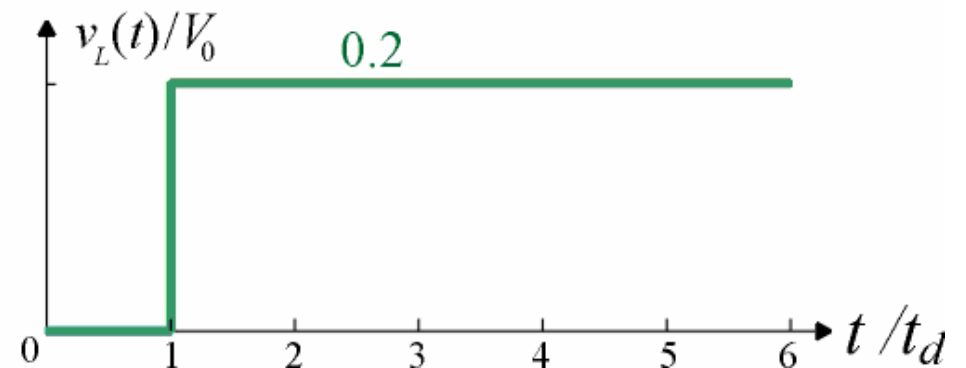
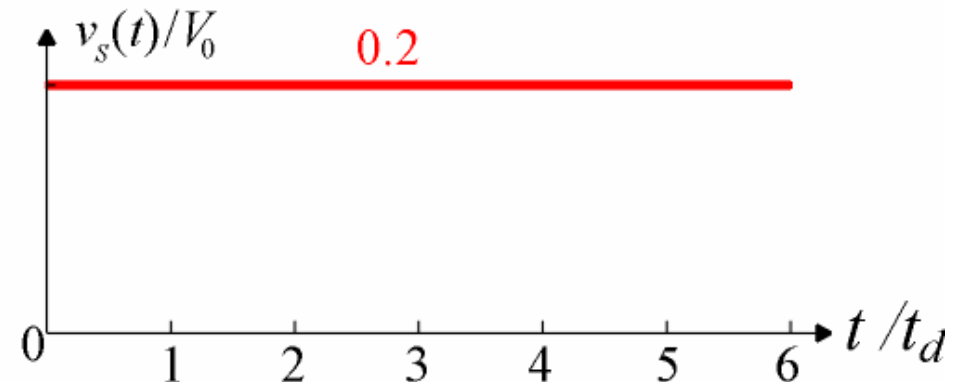
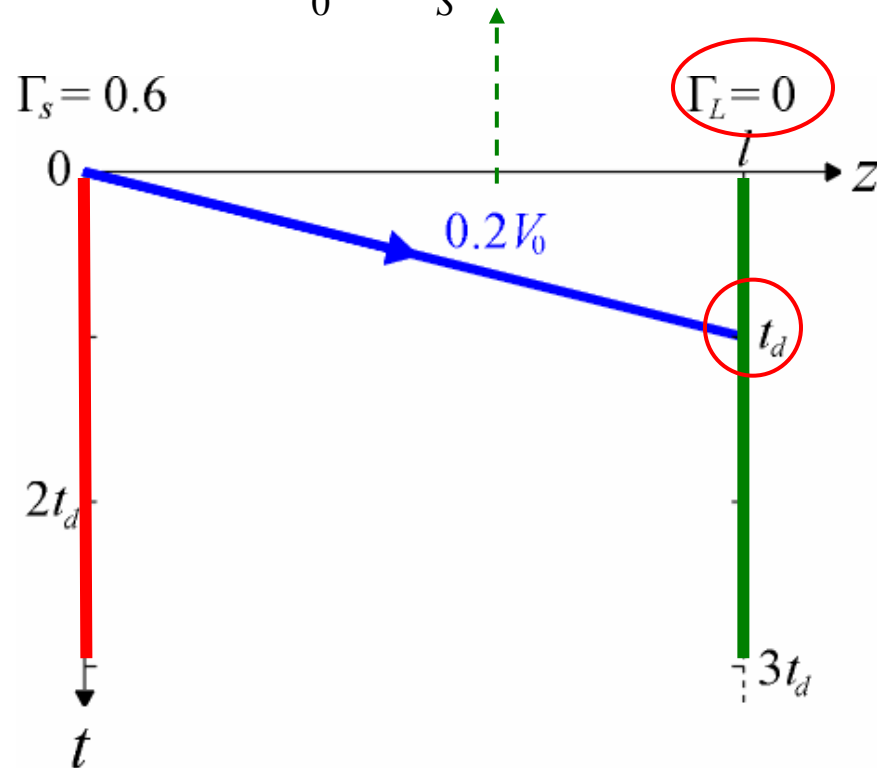


$$\Gamma_s = \frac{R_s - Z_0}{R_s + Z_0} = 0.6$$

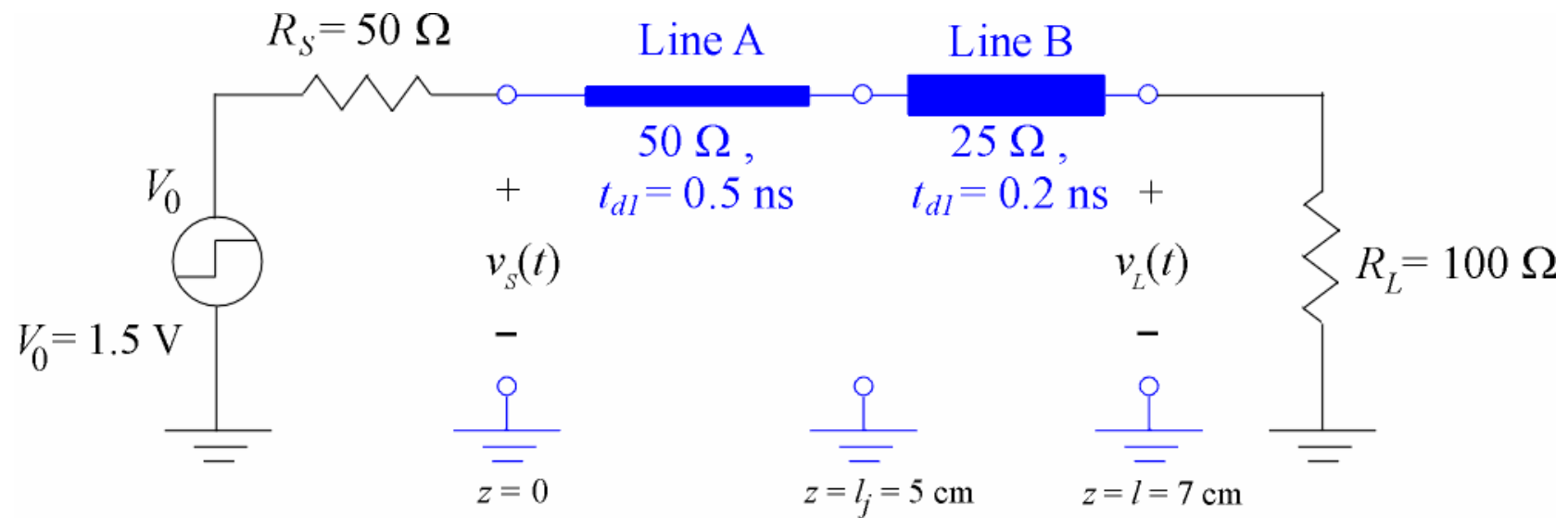
### Example 3-3: Single TX line with matched termination-2

$$V_1^+ = \frac{Z_0}{Z_0 + R_S} V_0 = 0.2V_0$$

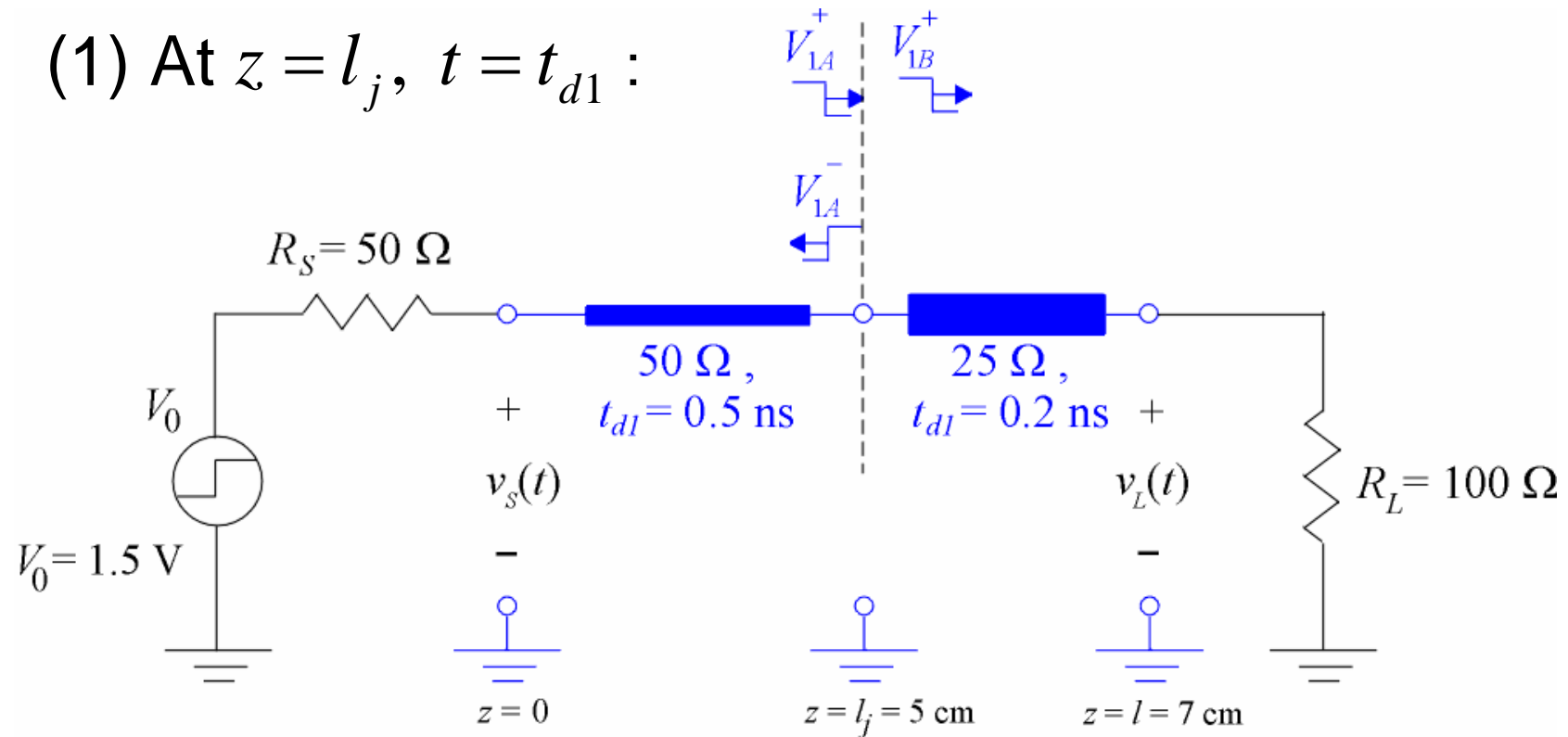
No overshooting &  
ringing (good!)



### Example 3-4: Cascaded TX line-1



### Example 3-4: Cascaded TX line-2



$$\text{BC: } v_{1A}^+(l_j, t_{d1}) + v_{1A}^-(l_j, t_{d1}) = v_{1B}^+(l_j, t_{d1})$$



### Example 3-4: Cascaded TX line-3

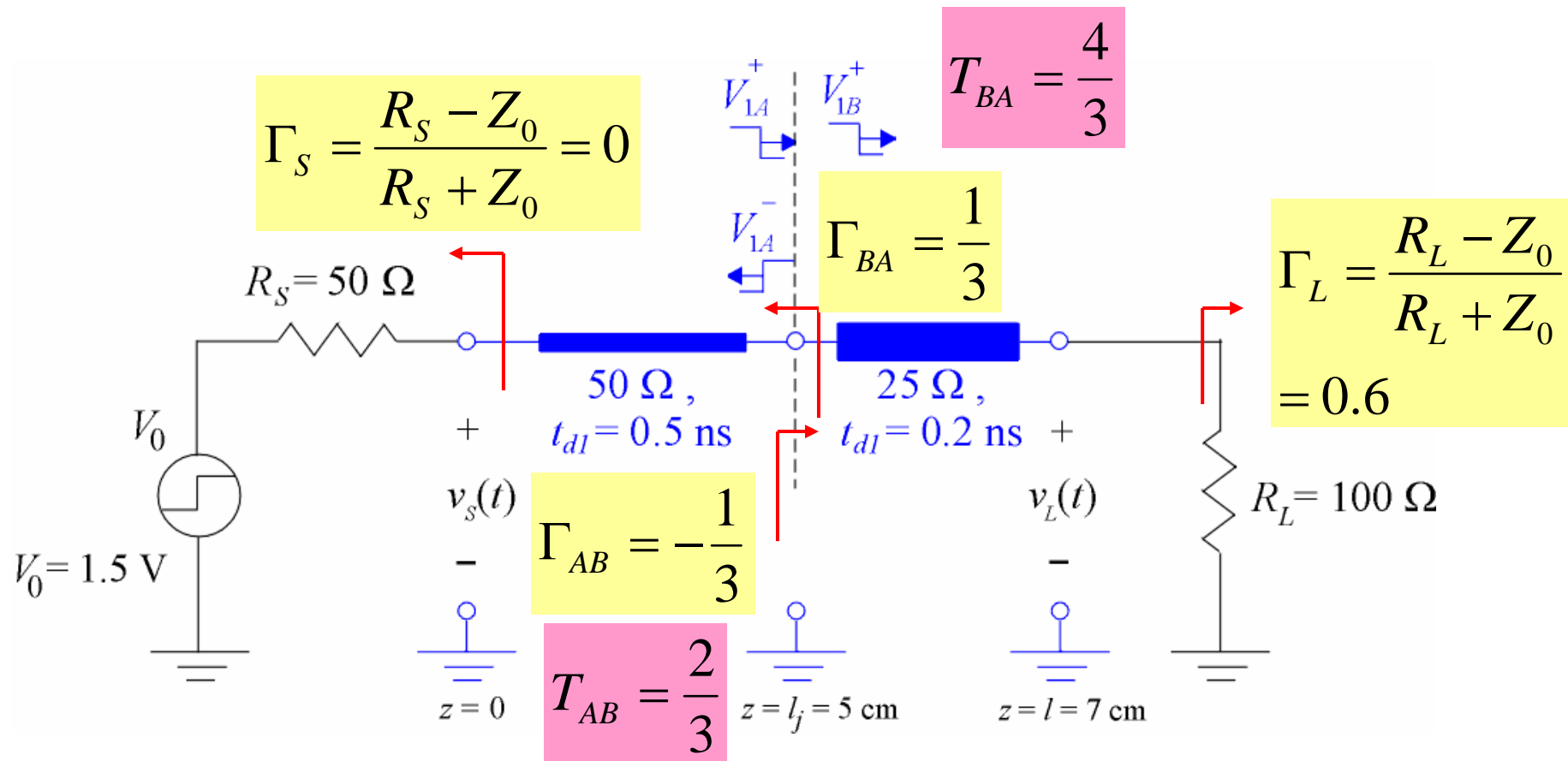
$$v_{1A}^+(l_j, t_{d1}) + v_{1A}^-(l_j, t_{d1}) = v_{1B}^+(l_j, t_{d1})$$

$$\Rightarrow 1 + \frac{v_{1A}^-(l_j, t_{d1})}{v_{1A}^+(l_j, t_{d1})} = \frac{v_{1B}^+(l_j, t_{d1})}{v_{1A}^+(l_j, t_{d1})},$$

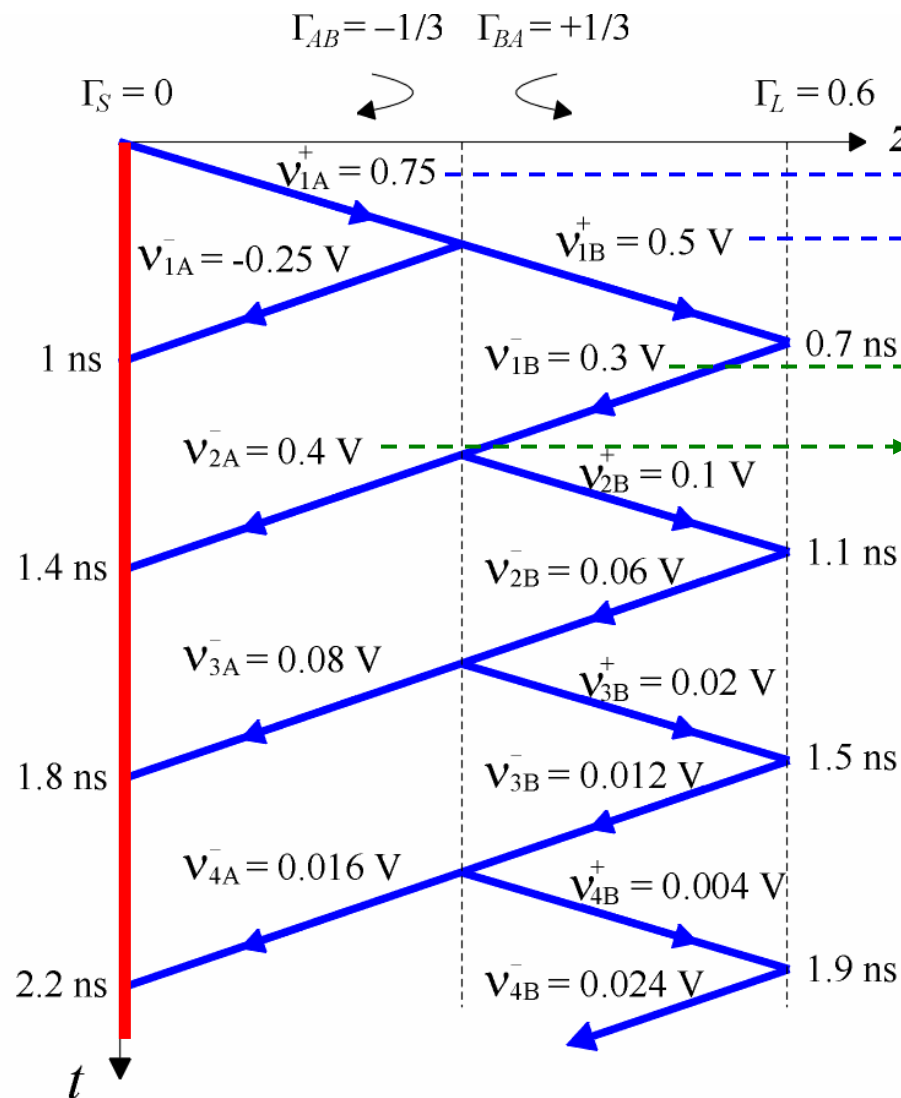
$$\Rightarrow 1 + \Gamma_{AB} = T_{AB} \quad \dots \text{Transmission coefficient}$$



### Example 3-4: Cascaded TX line-4



## Example 3-4: Cascaded TX line-5

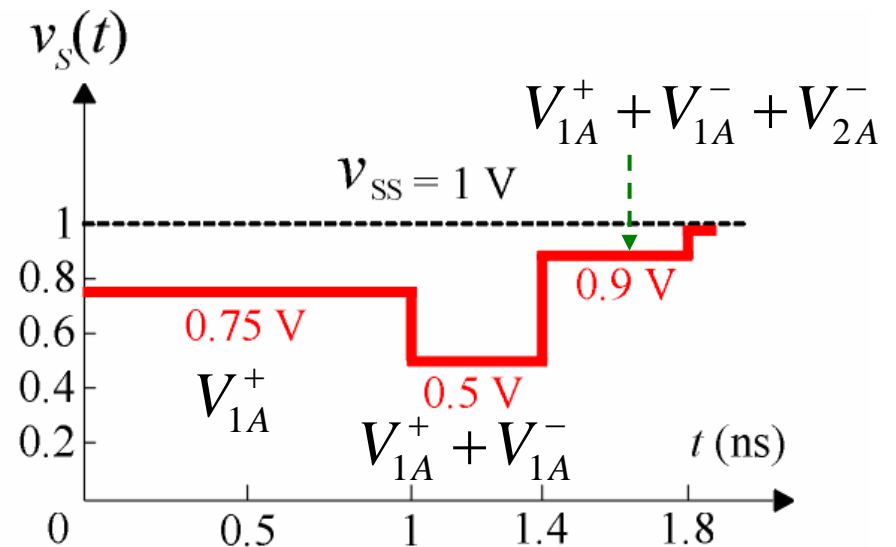


$$V_{1A}^+ = \frac{Z_{0A}}{Z_{0A} + R_S} V_0 = 0.5V_0 = 0.75 \text{ V}$$

$$V_{1B}^+ = (0.75 \text{ V}) \cdot T_{AB}$$

$$V_{1B}^- = (0.5 \text{ V}) \cdot \Gamma_L$$

$$V_{2A}^+ = (0.3 \text{ V}) \cdot T_{BA}$$



## Example 3-4: Cascaded TX line-6

