Final Exam

1) The Laplace transform of a function f(t) is derived by an integral

$$F(s) = L\{f(t)\} \equiv \int_{0^{-}}^{\infty} f(t)e^{-st}dt$$
.

- 1A) (5%) Derive F(s) and indicate the valid region on the s-plane if $f(t) = e^{-\alpha t}u(t)$, where $\alpha > 0$ and u(t) stands for the unit-step function.
- **(1B)** (10%) Perform <u>partial fraction expansion</u> for a rational function

$$F(s) = \frac{s}{\left[s - (-\alpha + j\beta)\right] \times \left[s - (-\alpha - j\beta)\right]} \times \frac{1}{s + \alpha}$$

where α , $\beta > 0$. Represent all the complex expansion coefficients in polar form (amplitude and phase).

- (10%) Follow Problem 1B, carry out inverse Laplace transform to get f(t).
- 2) Consider a circuit shown in Figure 1.



Figure 1.

- 2A) (5%) Let the input and output be v_1 and v_2 , respectively. What type of filters does this circuit correspond to? (*Hint*: Check the v_2/v_1 values at $\omega \rightarrow 0$ and $\omega \rightarrow \infty$.)
- 2B) (5%) Plot the s-domain circuit if no initial energy stored in the capacitor or the inductor.
- **2C)** (5%) Derive the <u>transfer function</u> of the circuit $H(s) \equiv V_2(s)/V_1(s)$.

- 2D) (5%) Roughly sketch the transmission spectrum $|H(j\omega)|$. (Note that this curve has to be consistent with your answer to Problem 2A.)
- 2E) (10%) Derive the <u>transfer function</u> of the circuit $H(s) \equiv V_2(s)/V_1(s)$ if a <u>load</u> resistance of $R_L = \frac{1}{4}\Omega$ is connected to the output port (the two terminals of v_2).
- 2F) (10%) Follow Problem 2E, denote the positions of all the pole(s) and zero(es) of H(s) by cross(es) "×" and open circle(s) "o", respectively. Estimate what type of filters does this loaded circuit correspond to accordingly?
- 2G) (5%) Follow Problem 2E, what are the values of peak transmission T_{max} and center frequency ω_c if they are defined by $|H(j\omega)| \le |H(j\omega_c)| = T_{\text{max}}$? (*Hint*: The curve

$$\frac{\beta\omega}{\sqrt{\left(\omega^2 - \omega_0^2\right)^2 + \left(\beta\omega\right)^2}} \quad \text{corresponds to} \quad T_{\text{max}} = 1, \quad \omega_c = \omega_0.$$

- 2H) (15%) Follow Problem 2E, derive the time-domain <u>output voltage</u> $v_2(t)$ if the loaded circuit is driven by an ideal voltage <u>source</u> of $v_g(t) = e^{-0.25t}u(t)$. (*Hint*: $v_g(t) = v_1(t)$, and $V_2(s) = H(s) \times V_g(s)$ in this case. You can borrow the results in Problems 1B and 1C with minor adjustment.)
- 2I) (15%) For the "unloaded" circuit shown in Figure 1, calculate the 2×2 transmission matrix [A] for the two-port circuit. Denote the physical units whenever exist. (*Hint*: [A] is defined by $\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} a_{11} & -a_{12} \\ a_{21} & -a_{22} \end{bmatrix} \times \begin{bmatrix} V_2 \\ I_2 \end{bmatrix}$. The matrix elements can be functions of *s*.)
- 2J) (Bonus 10 points) The configuration of Problem 2H is a special case of "terminated two-port circuit" where the source impedance $Z_g = 0$. List the equations that can be used in solving $\{V_1, I_1, V_2, I_2\}$ in the s-domain. Show that the $V_2(s)$ derived by solving the listed simultaneous equations is consistent with that obtained by $H(s) \times V_g(s)$.